



**VLSI AND SOFTWARE RADIOS**  
**By John Fakatselis**  
**Harris semi.**



## CONTRIBUTING ENABLING TECHNOLOGIES TO VLSI RADIOS.

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- **Process Technology**
- **Package Technology**
- **IC Simulation Tools**
- **System Simulation Tools**



## RADIO DESIGN EXAMPLE

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- **AN EXAMPLE OF A VLSI RADIO DESIGN:**
  - **WIRELESS LAN, BASED ON PROPOSED IEEE8021.11 SPEC.**
  - **RF, IF , MODEM DESIGN , AND PROTOCOL MAC.**



# Wireless LAN (WLAN) Markets

## Current Market

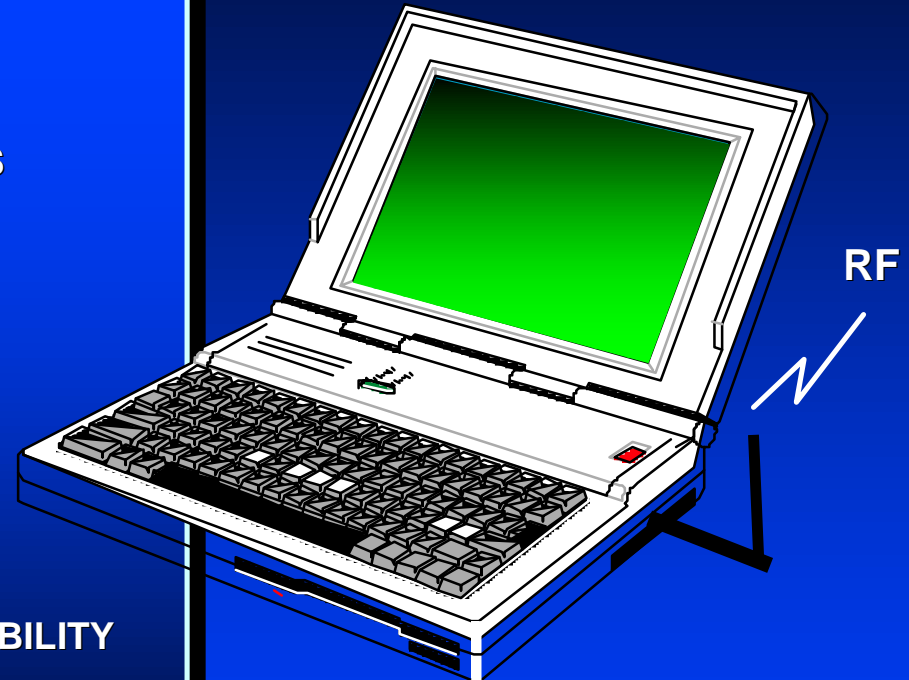
PRIMARILY AT 902 - 928MHz

- MIX OF SPREAD SPECTRUM TRANSCEIVERS WITHOUT STANDARDS
- NO COMPLIANT PROTOCOL STANDARDS
- NO INTEROPERABILITY OF EQUIPMENT
- LOW DATA RATES

## Emerging Market

2.4 - 2.5GHz

- IEEE 802.11 STANDARD FOR WLANS
- COMPLIANT STANDARDS FOR INTEROPERABILITY
- CONNECTIVITY
- SUPPORT HIGHER DATA RATES
- PCMCIA LAN CARD SOLUTION
- LOWER COST
- HIGHER PERFORMANCE
- SPEED, RANGE, BATTERY LIFE





# The 802.11 Wireless LAN Standard



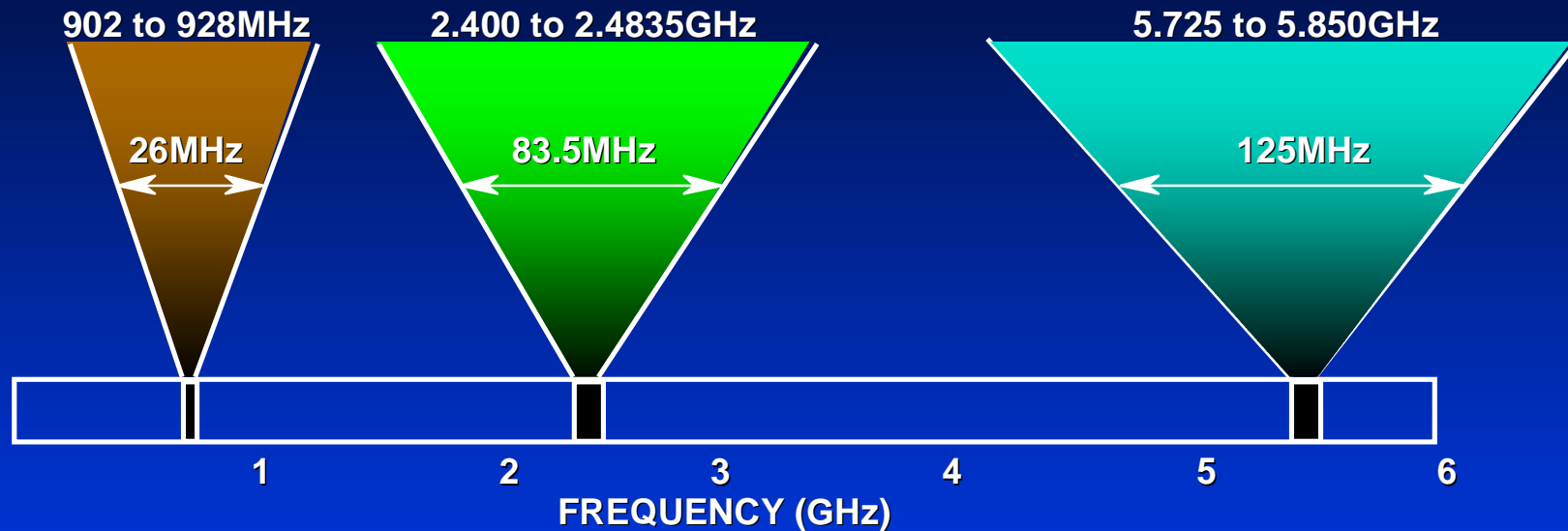
# IEEE 802.11 WLAN Standard

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- IEEE WLAN WORKING GROUP WITH GLOBAL REPRESENTATION
- STANDARD FOR WLANS WITH A COMMON MULTI-USER MEDIA
- GLOBAL EQUIPMENT INTEROPERABILITY
- FREQUENCY: 2.4GHz ISM BAND (83.5MHz BW)
- MEDIUM TO HIGH DATA RATES
- APPLICATIONS RANGING FROM SMALL OFFICES TO INDUSTRIAL MANUFACTURING CAMPUSES
- SPREAD SPECTRUM TECHNOLOGY
  - Limits Transmitted Power Density
  - Provides a Robust Solution in a Multi-user Environment



# Industrial, Scientific and Medical (ISM) Bands



- UNLICENSED OPERATION GOVERNED BY FCC DOCUMENT 15.247, PART 15
- SPREAD SPECTRUM ALLOWED TO MINIMIZE INTERFERENCE
- 2.4GHz ISM BAND
  - More Bandwidth to Support Higher Data Rates and Number of Channels
  - Available Worldwide
  - Good Balance of Equipment Performance and Cost Compared with 5.725GHz Band
  - IEEE 802.11 Global WLAN Standard

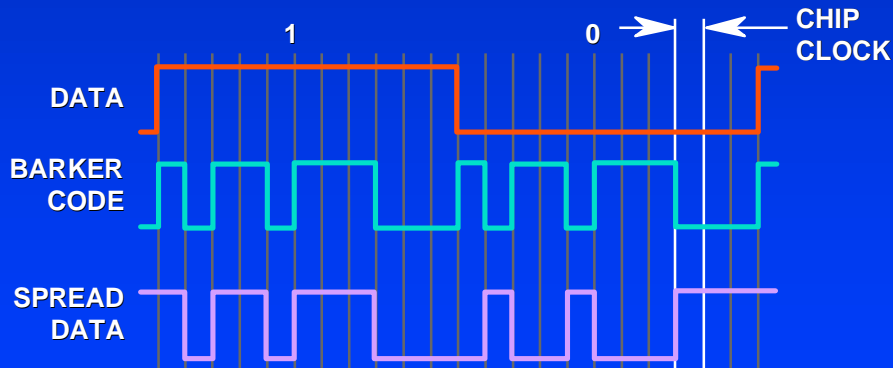
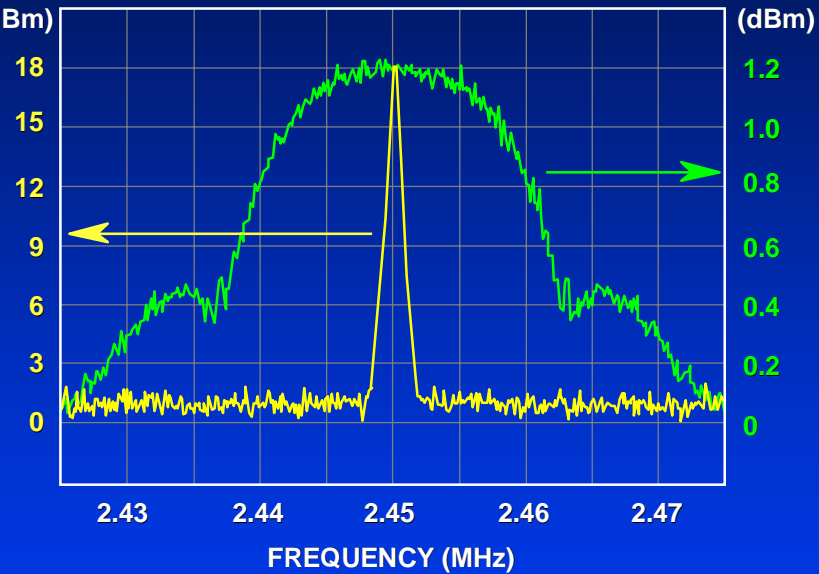


# Direct Sequence Spread Spectrum

- DATA SIGNAL SPREAD BY A PN CODE
- PROPERTIES OF PN CODE
- CHIP RATE
- DS PROCESSING GAIN
$$G_p \text{ (dB)} = 10 \text{LOG} \left( \frac{\text{CHIP RATE}}{\text{DATA RATE}} \right)$$
- PN CORRELATION AT RECEIVER
- PSK DATA MODULATION

CW SIGNAL  
AMPLITUDE  
(dBm)

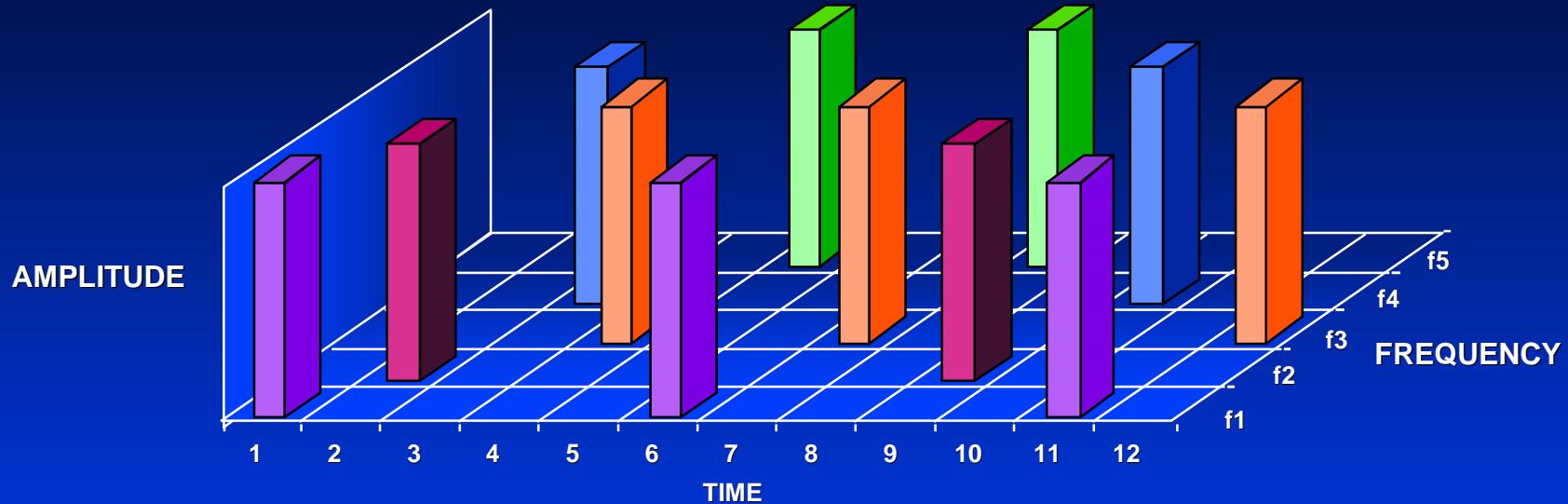
SPREAD SIGNAL  
AMPLITUDE  
(dBm)







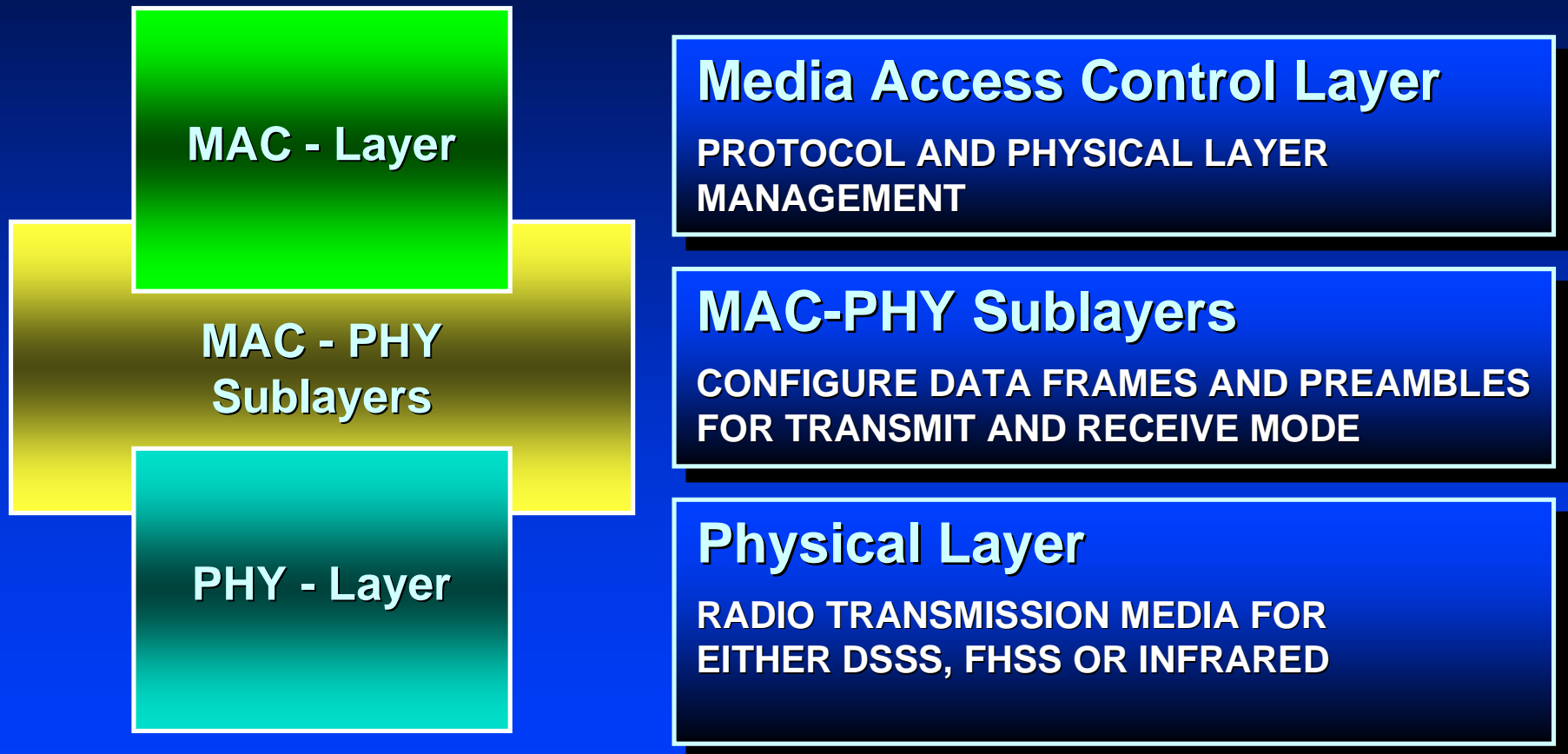
# Frequency Hopping Spread Spectrum



- FSK DATA MODULATION
- PERIODIC CHANGES IN THE CARRIER FREQUENCY SPREADS THE SIGNAL
- CARRIER FREQUENCY CHANGES AT A SPECIFIED HOP RATE
- CARRIER FREQUENCY HOPS AFTER A PRESCRIBED TIME
- TOTAL SYSTEM BANDWIDTH INCLUDES ALL OF THE CHANNEL FREQUENCIES USED IN HOPPING



# 802.11 Protocol Layers





# 802.11 DSSS Physical Layer

MODULATION:	DPSK	
PN CODE:	11-BIT BARKER	
DATA RATE:	1Mbps/DBPSK, 2Mbps/DQPSK	
CHIPPING RATE:	11Mcps	
TRANSMIT POWER:	USA: 1W MAX.	FCC 15.247
	EUROPE: 100mW (EIRP)	ETSI 300-328
	JAPAN: 10mW / MHz	MPT ORD. 78
TRANSMIT FREQUENCY TOLERANCE:	±25ppm MAX.	
CHIP CLOCK ACCURACY:	±25ppm MAX.	
RECEIVER INPUT LEVEL SENSE:	-80dBm $8 \times 10^{-2}$	FER (FRAME ERROR RATE)
RECEIVER ENERGY DETECT:	20µs MAX.	
PREAMBLE LENGTH:	144 SYMBOLS	
DATA PACKET SIZE:	2048 BYTES MAX.	
ENVIRONMENT:	PACKET BURST DATA	

The logo features a black horizontal bar with a white 'PRISM™' text on the left and a stylized antenna symbol on the right. The antenna symbol consists of three concentric curved lines radiating from a point on the right side of the bar. A small 'TM' trademark symbol is located at the bottom right of the antenna symbol.

**PRISM™ Chipset**



# WLAN Radio Requirements

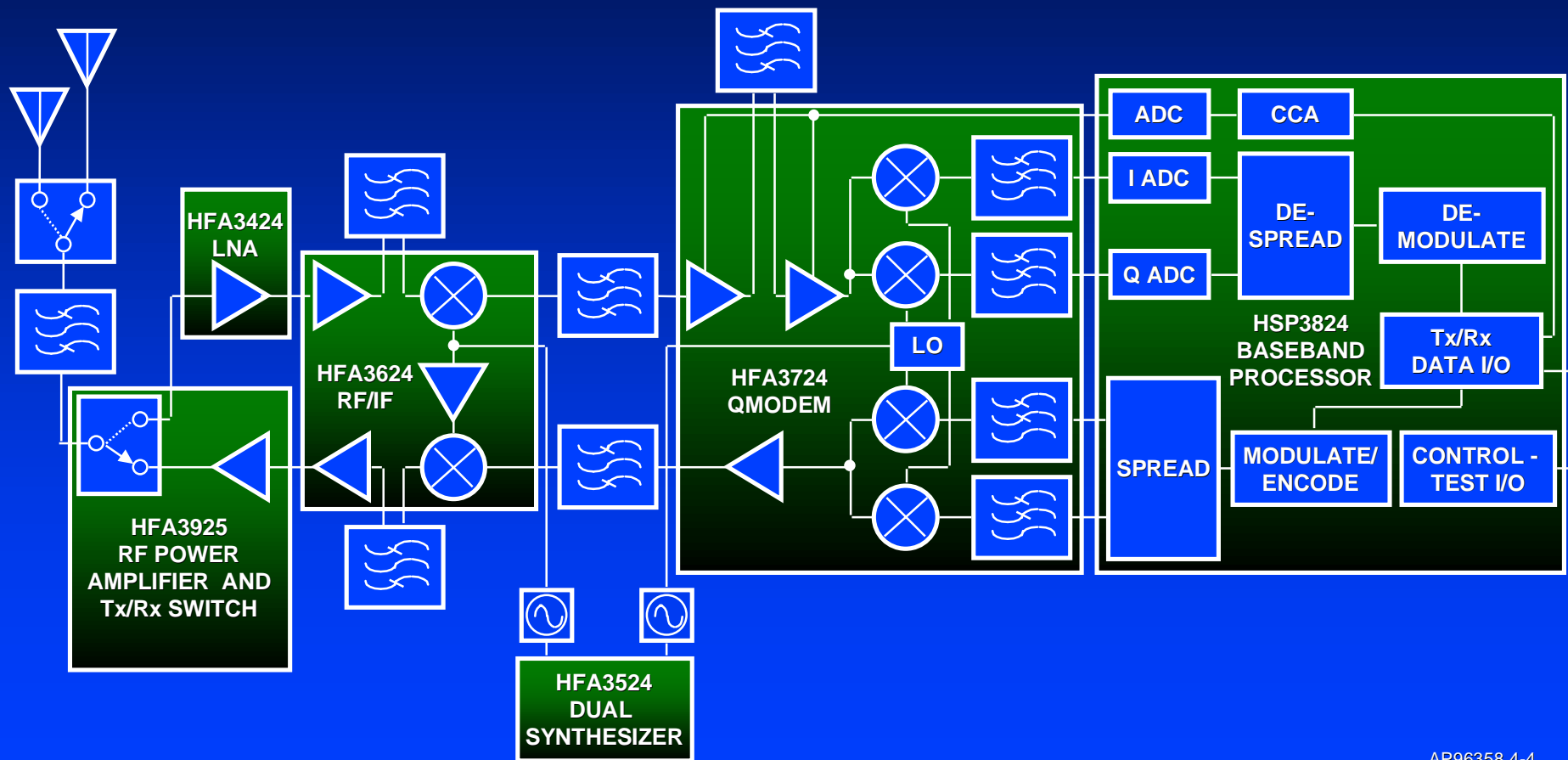


- LOW COST
- HIGH DATA RATE AND HIGH THROUGHPUT
- PCMCIA COMPATIBLE
- LONG RANGE
- MOBILITY WITH ROAMING CAPABILITY
- LOW VOLTAGE, LOW POWER, BATTERY OPERATED
- HIGH LEVEL OF INTEGRATION



# PRISM™ “Antenna to Bits™”

## A Complete DS Spread Spectrum Radio Chipset





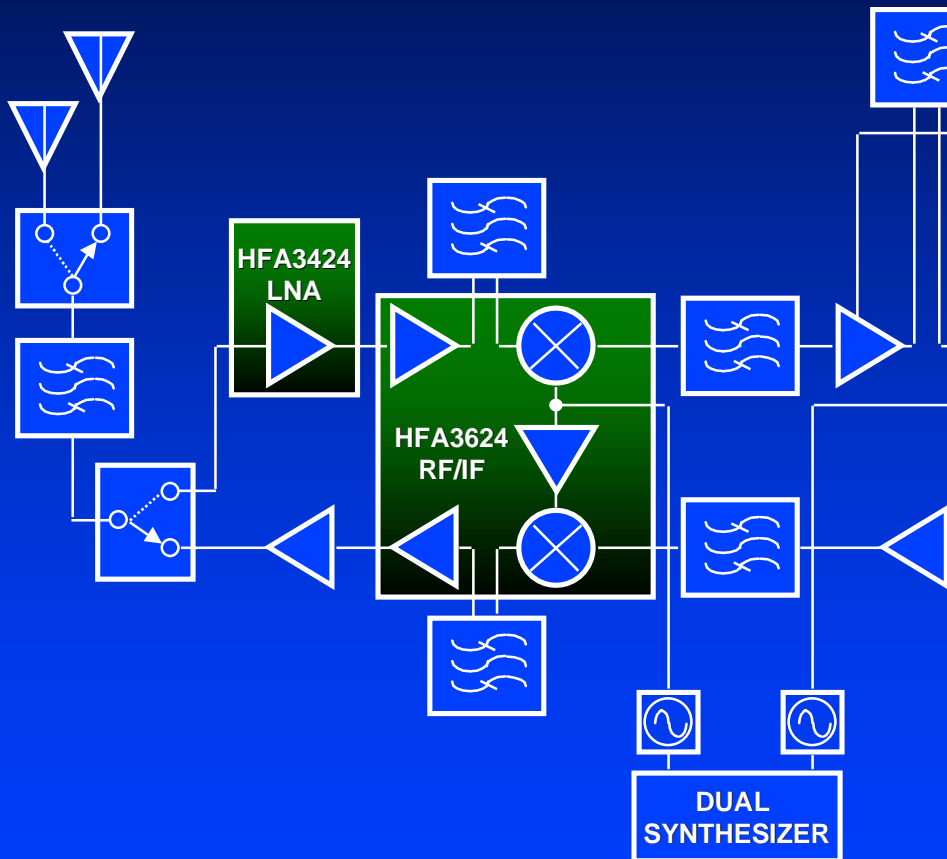
# RF IC Partitioning Criteria

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- **FREQUENCY**
- **ANALOG / DIGITAL SIGNALS**
- **PROCESS TECHNOLOGY**
- **ISOLATION REQUIREMENTS**
- **POWER LEVELS**
- **PACKAGING**
- **EXTERNAL COMPONENTS**



# RF to IF Conversion Issues

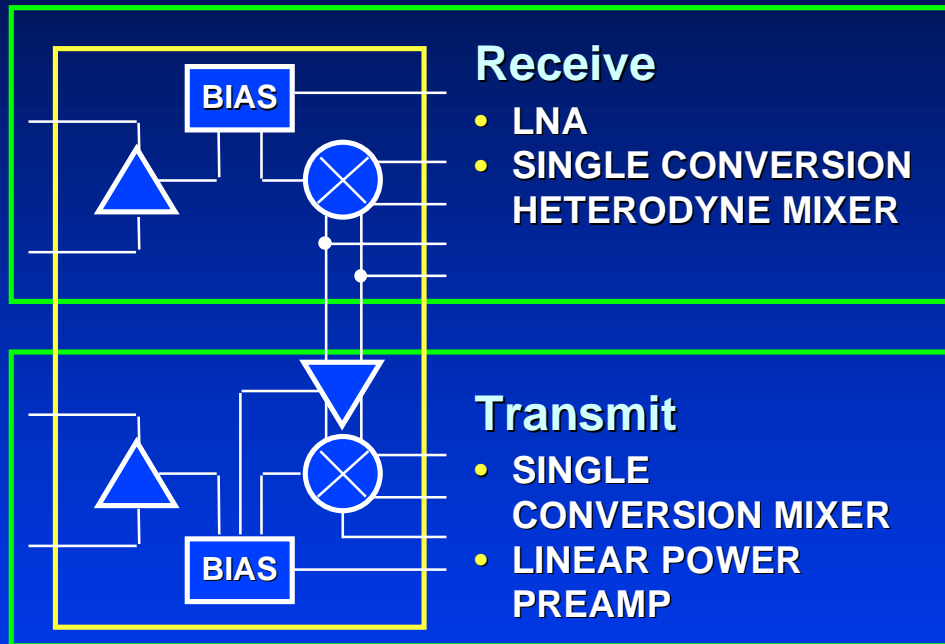


- NOISE FIGURE (NF)
- INPUT/OUTPUT 1dB COMPRESSION POINT (IP1dB/OP1dB)
- INPUT/OUTPUT THIRD ORDER INTERCEPT POINT (IP3<sub>i</sub>/ IP3<sub>o</sub>)
- IMPEDANCE MATCHING
- DYNAMIC RANGE
- SPURIOUS FREE DYNAMIC RANGE (SFDR)





# HFA3624 RF to IF Converter



- INTEGRATED RECEIVE / TRANSMIT FRONT END
- 2.4 TO 2.5GHz RF FREQUENCY RANGE
- 10 TO 400MHz IF OPERATION
- SINGLE SUPPLY 2.7V TO 5.5V
- ALL RF I/Os MATCHED TO 50Ω
- INDEPENDENT RECEIVE / TRANSMIT POWER ENABLE

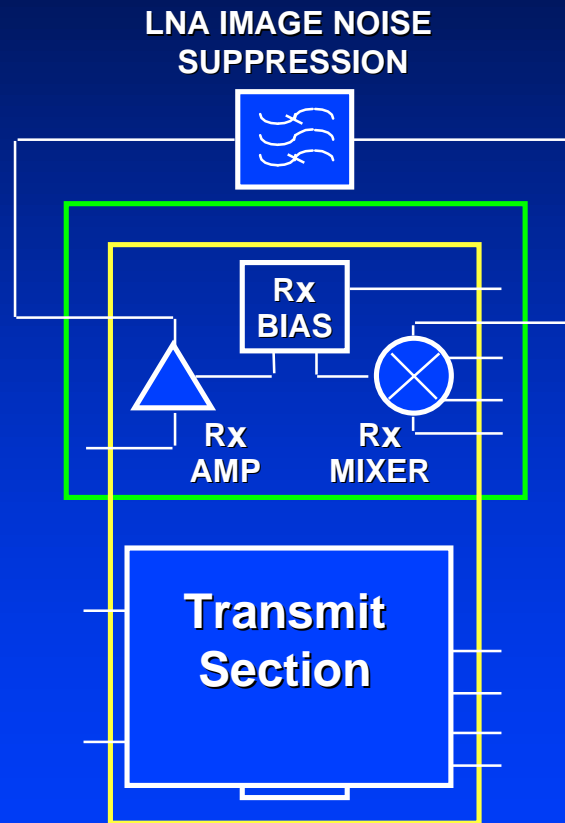
## 28-PIN SSOP

- 150mil WIDE
- 0.08 SQUARE INCHES

Rx	Tx	POWER CONSUMPTION
0	0	0.3μA
0	1	48mA
1	0	18mA



# HFA3624 Receive Section



## LNA @ 2.5GHz

- POWER GAIN ..... 15.6dB
- OP1dB ..... 5.5dBm
- NF ..... 3.8dB

## Mixer (LO = 2.15GHz)

- POWER CONVERSION GAIN ..... 3.0dB
- IP<sub>3O</sub> ..... 4.0dB
- NF ..... 12.0dB

NF = 10LOG(F) = NOISE FIGURE  
WHERE (F) = NOISE FACTOR  
F = SNR<sub>i</sub> / SNR<sub>o</sub>



# HFA3624 RF Front End Cascaded Performance



## Cascaded Performance Data

- NF..... 4.5dB
- IP1dB ..... -28dB
- IP3<sub>1</sub>..... -18dB

## Cascaded Performance Data

$$F_C = F_1 + \frac{F_2 - 1}{G_1} + \frac{F_3 - 1}{G_1 G_2} + \dots$$

- F<sub>C</sub> = CASCADED NOISE FACTOR
- G<sub>n</sub> = GAIN OF STAGE n

## System Performance

- DYNAMIC RANGE (IP1dB - MDS)
- SPURIOUS FREE DYNAMIC RANGE  
2/3(IP3<sub>1</sub> - MDS)

## F<sub>c</sub> Effect on Sensitivity (MDS)

- MDS = F<sub>C</sub>(kTB) (SNR)<sub>o</sub>

where: MDS = Minimum Discernible Signal

k = Boltzmann's Constant

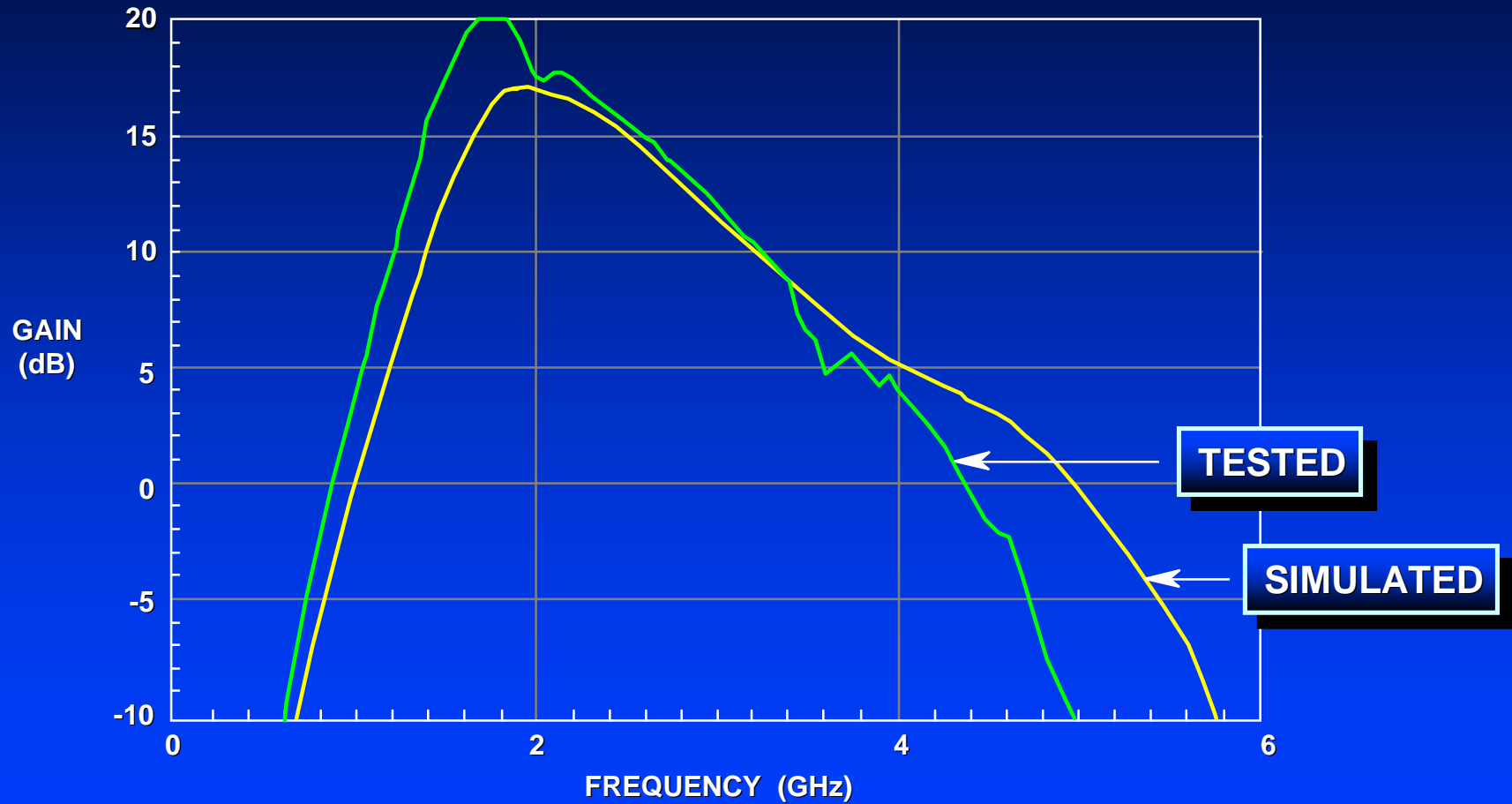
T = Absolute Temperature in Kelvin

B = Bandwidth

(SNR)<sub>o</sub> = Output Signal-to-Noise Ratio

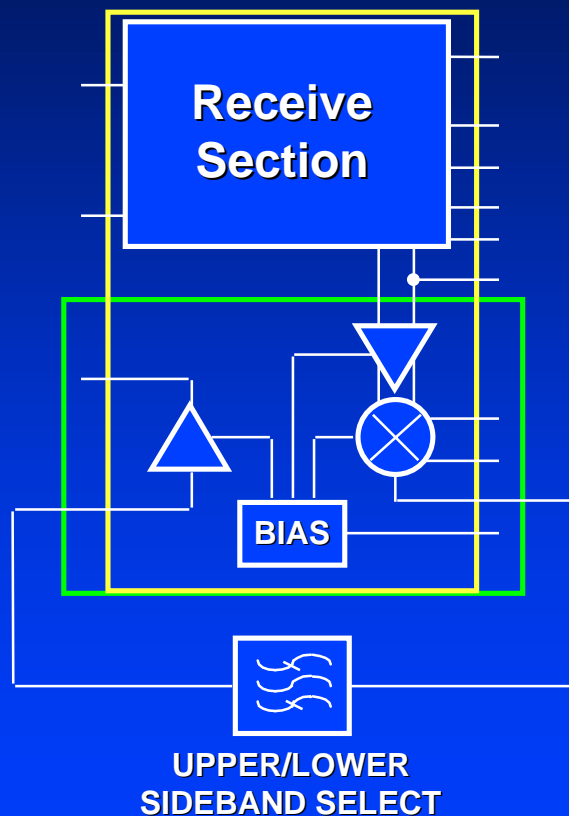


# HFA3624 Receive Amplifier Power Gain (S21)





# HFA3624 Transmit Section



## Mixer (Upconverter)

- POWER CONVERSION GAIN . . . . . 5.8dB
- LO LEAKAGE . . . . . -20.0dB

## Pre-Amp

- POWER GAIN . . . . . 13.0dB
- OP1dB . . . . . +7.5dBm

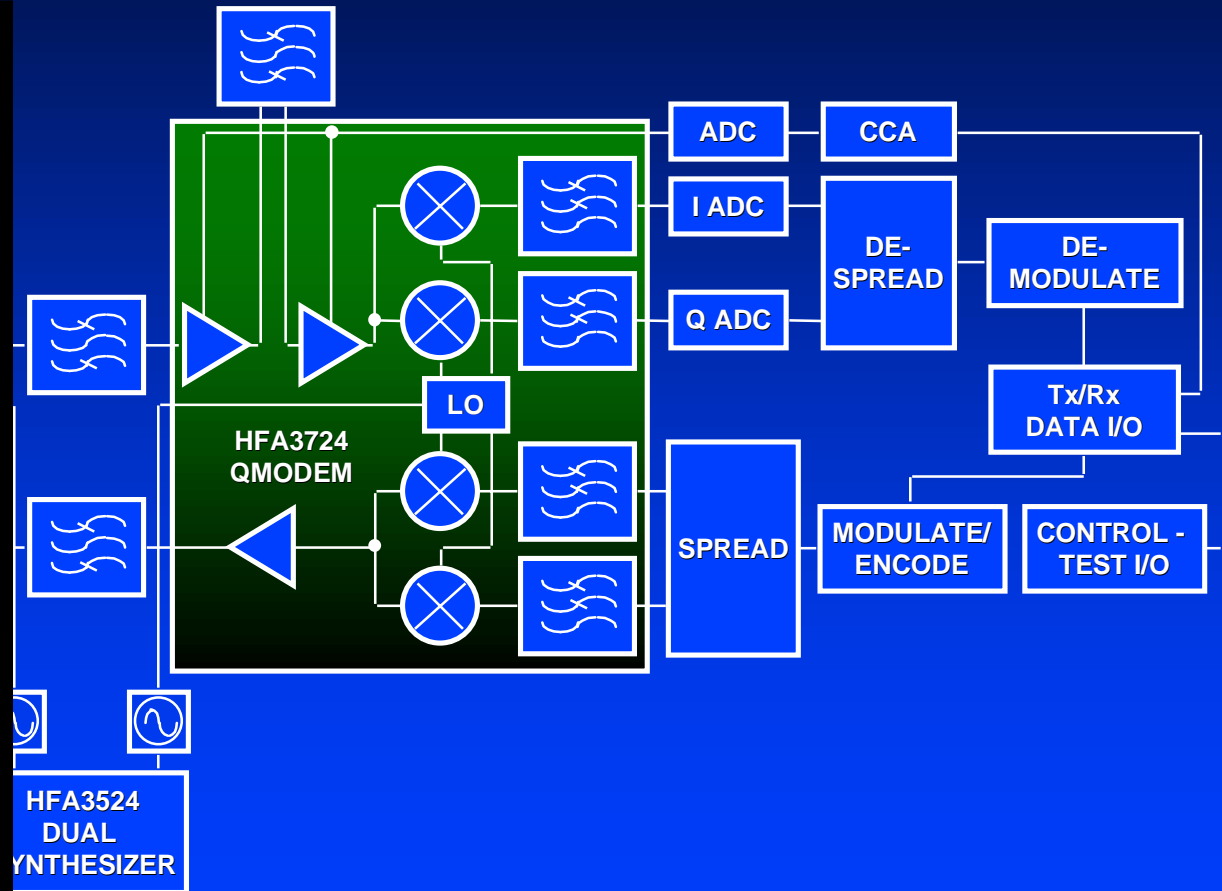
## Cascaded Performance

- POWER GAIN . . . . . 17.8dB
- OP1dB . . . . . 0.5dBm
- LO LEAKAGE . . . . . -5.0dBm



# IF to Baseband Conversion Issues

- PRIMARY SOURCE OF GAIN
  - Limiter or AGC
- RSSI TO MEASURE INPUT SIGNAL
- COMPLEX DOWN CONVERSION TO BASEBAND
- QUADRATURE LO GENERATION
- FILTERING (BASEBAND)
  - Anti-aliasing on Receive
  - Pulse Shaping on Transmit
- UP CONVERSION / MODULATION
- POWER SAVING FEATURES





# HFA3724 Quadrature IF Modulator / Demodulator

## (A) Limiter and RSSI

- IF UP TO 400MHz
- 84dB GAIN
- 70dB RSSI DYNAMIC RANGE

## (B) Demodulator

- DOWN CONVERTS 10 TO 400MHz IF TO BASEBAND
- IN PHASE (I) AND QUADRATURE (Q)

## (C) Filter

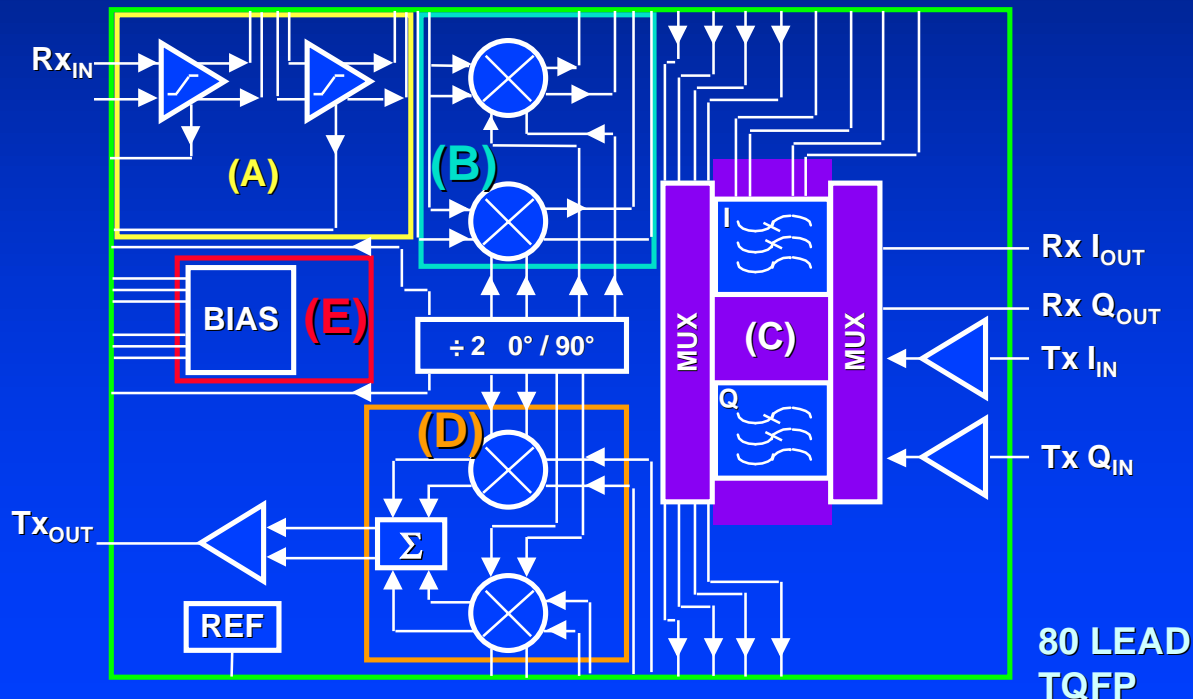
- TWO LOW PASS FILTERS (I AND Q)
- DIGITALLY SELECTABLE FILTER CUTOFF
- MULTIPLEXED RECEIVE / TRANSMIT

## (D) Modulator

- UP CONVERT AND PHASE MODULATES I AND Q TO 300MHz

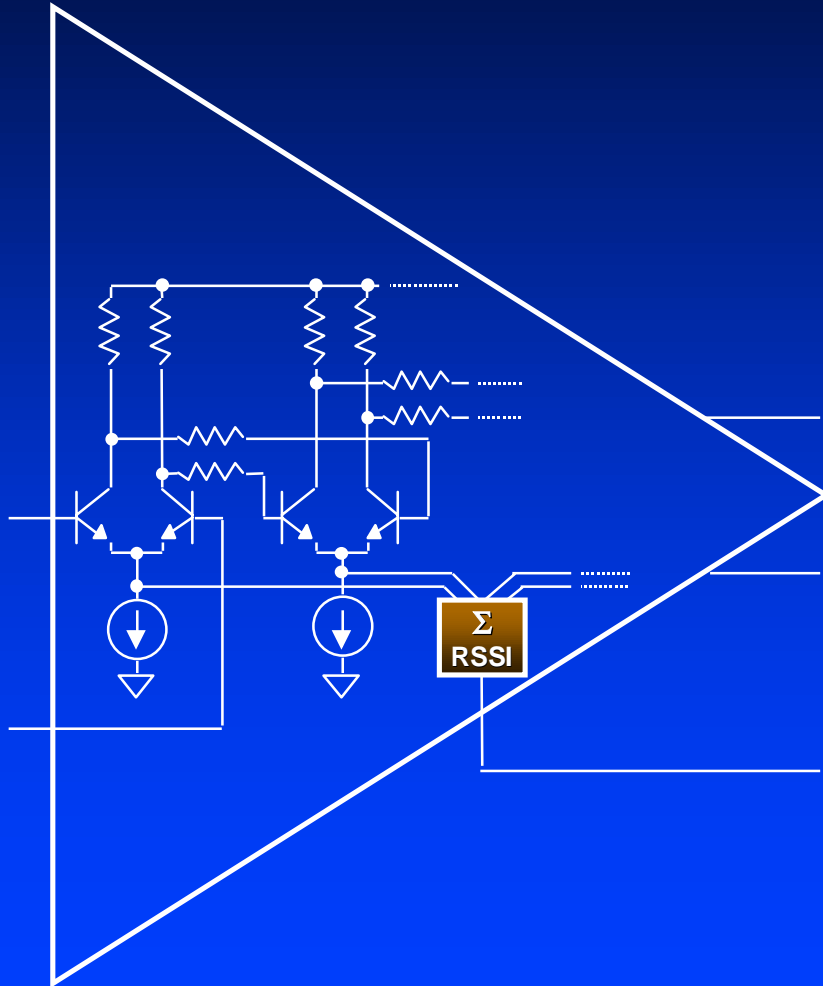
## (E) Power

- POWER CONTROL INCORPORATING SLEEP MODE
- HALF DUPLEX OPERATION





# HFA3724 Limiter and RSSI



## Limiter

- TWO INDEPENDENT AMPLIFIERS
- GAIN (EACH AMP)  
FROM 10 TO 400MHz ..... 42dB
- LIMITER -3dB  
SENSITIVITY..... -84dBm
- NF (FOR 250Ω SINGLE ENDED  
INPUT IMPEDANCE) ..... 6dB

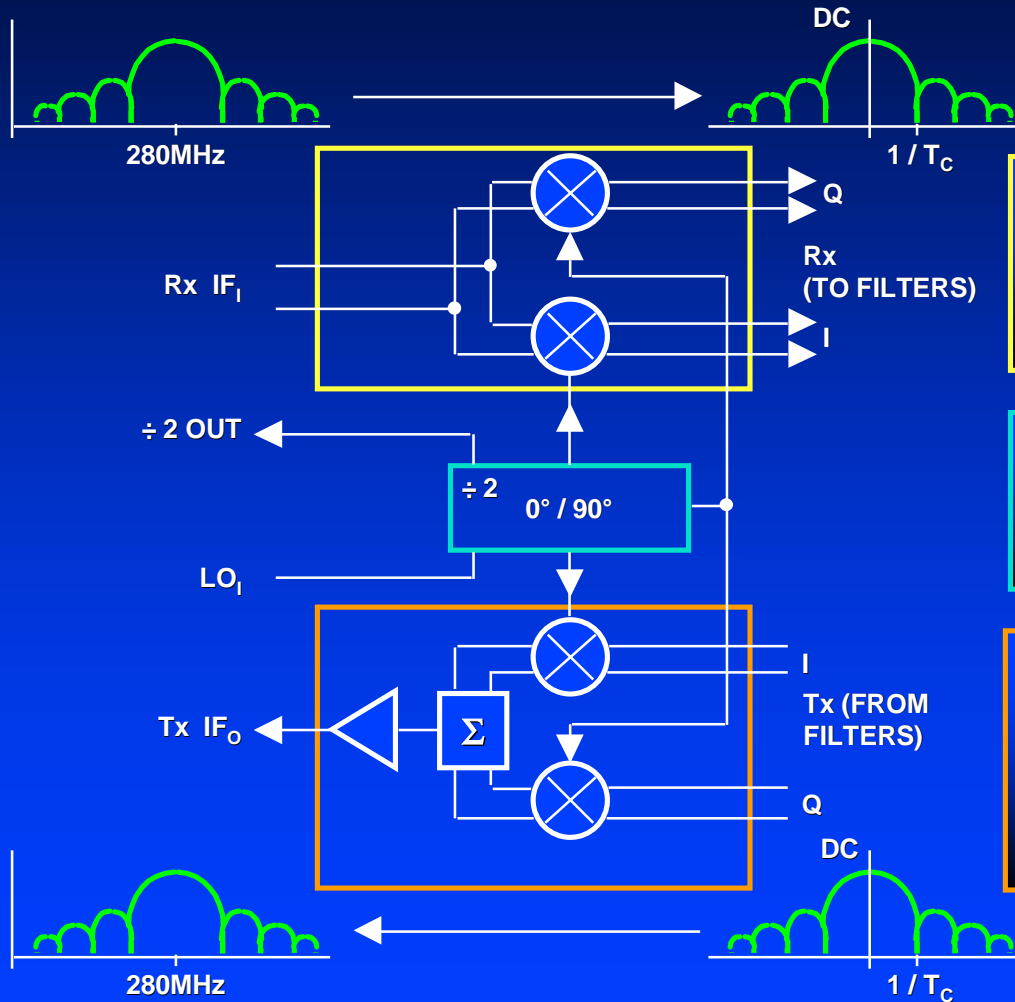
## RSSI

- SENSITIVITY..... -84dBm
- DYNAMIC RANGE ..... 70dB
- RSSI SLOPE ..... 5μA/dB





# HFA3724 Q-Modem



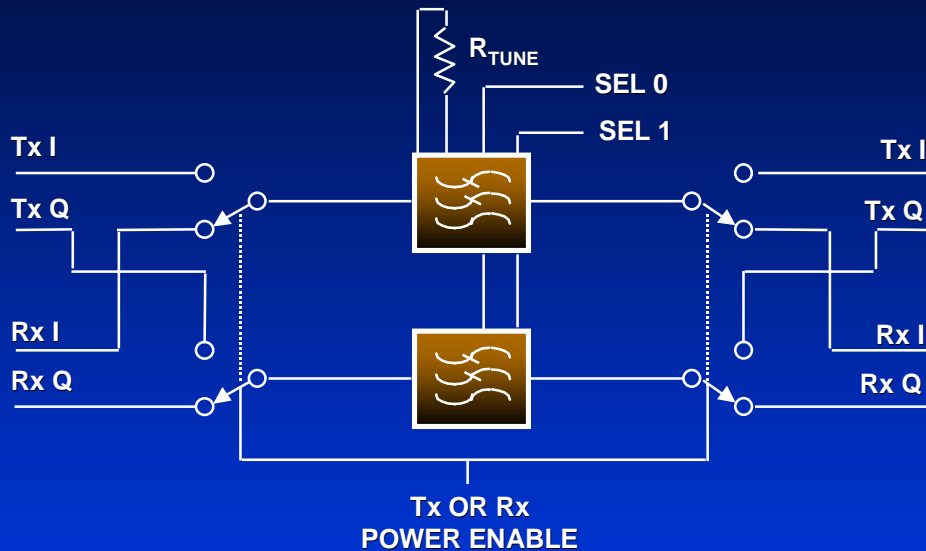
- ### Quadrature Downconverter
- INPUT FREQUENCY RESPONSE . . . . . 400MHz
  - DIFFERENTIAL VOLTAGE GAIN . . . . . 8dB
  - PHASE MATCH . . . . .  $\pm 2^\circ$
  - AMPLITUDE MATCH . . . . .  $\pm 0.5dB$

- ### Quadrature LO Generator
- INTERNAL DIVIDE BY 2 FLIP-FLOP WITH DUTY CYCLE COMPENSATION
  - $50\Omega$  DIVIDED BUFFER OUTPUT AVAILABLE

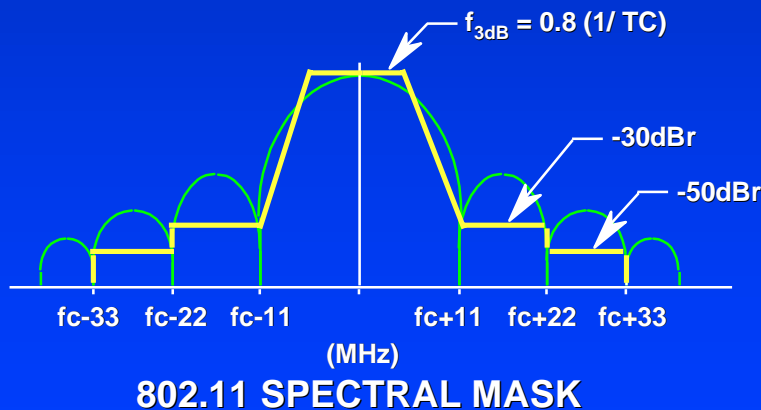
- ### Quadrature Upconverter
- PHASE MATCH . . . . .  $\pm 2^\circ$
  - AMPLITUDE MATCH . . . . .  $\pm 0.5dB$
  - LO LEAKAGE . . . . .  $-30dBc$
  - MINIMUM SIDEBAND SUPPRESSION (400MHz) . . . . .  $-26dBc$



# HFA3724 Baseband Filters



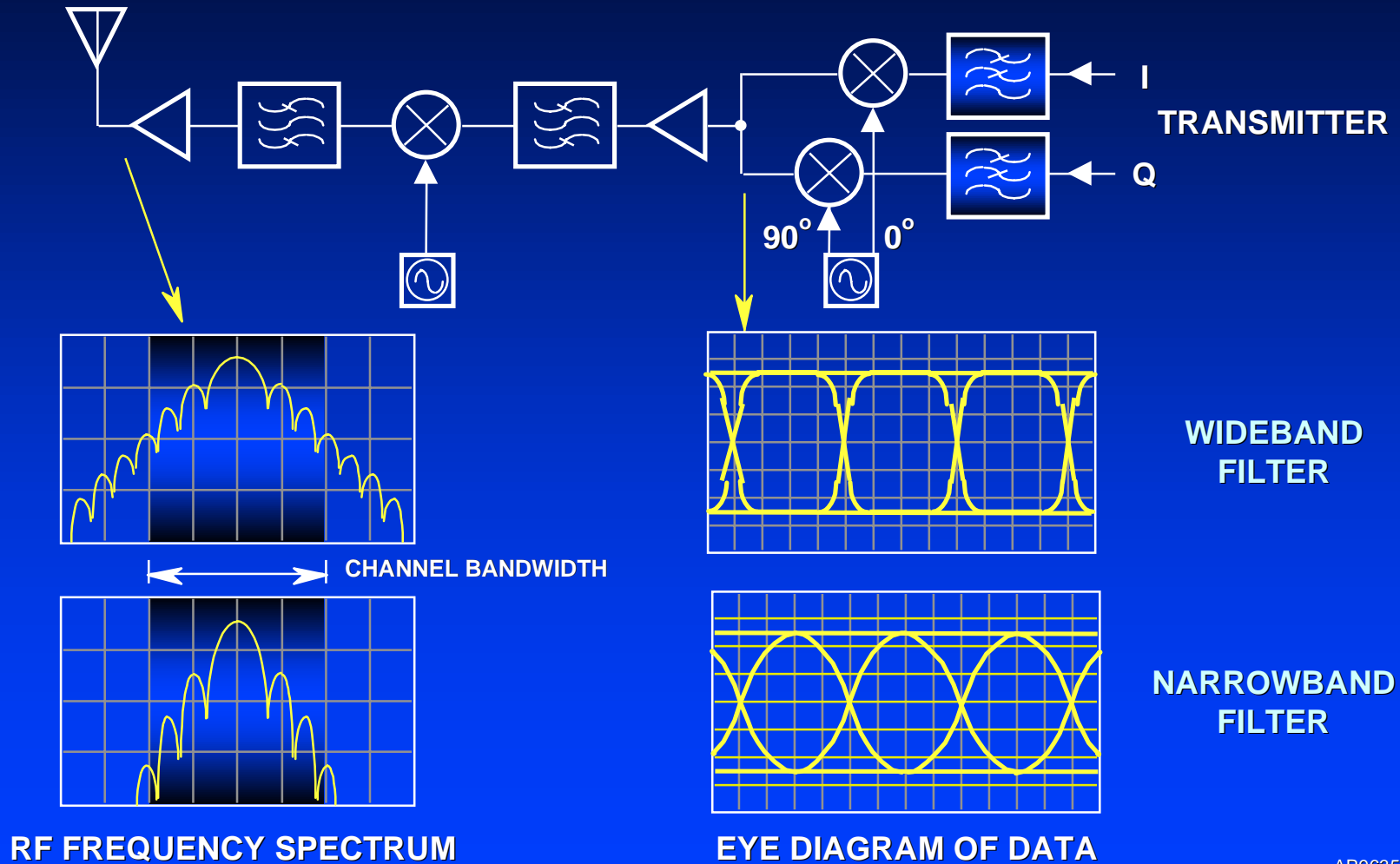
- 5 - POLE BUTTERWORTH FILTERS
- MULTIPLEXED FILTERS FOR HALF DUPLEX OPERATION
  - Pulse Shaping on Transmit
  - Anti-aliasing on Receive
- FOUR SELECTABLE CUTOFF
- FREQUENCIES: 2.2MHz, 4.4MHz, 8.8MHz, 17.6MHz
- $\pm 20\%$  FINE TUNING WITH EXTERNAL RESISTOR



- TRANSMIT FILTERING REDUCES SPECTRAL PRODUCTS TO IEEE802.11 SPECIFIED LEVELS
- TUNING RESISTOR COMPENSATES FOR SPECTRAL RE-GROWTH



# Adjacent Channel Interference Versus Intersymbol Interference

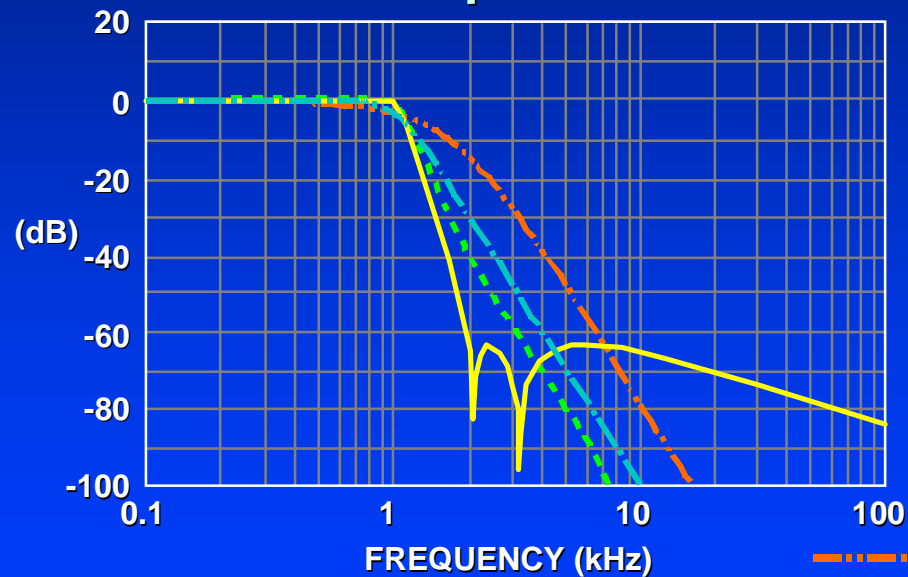




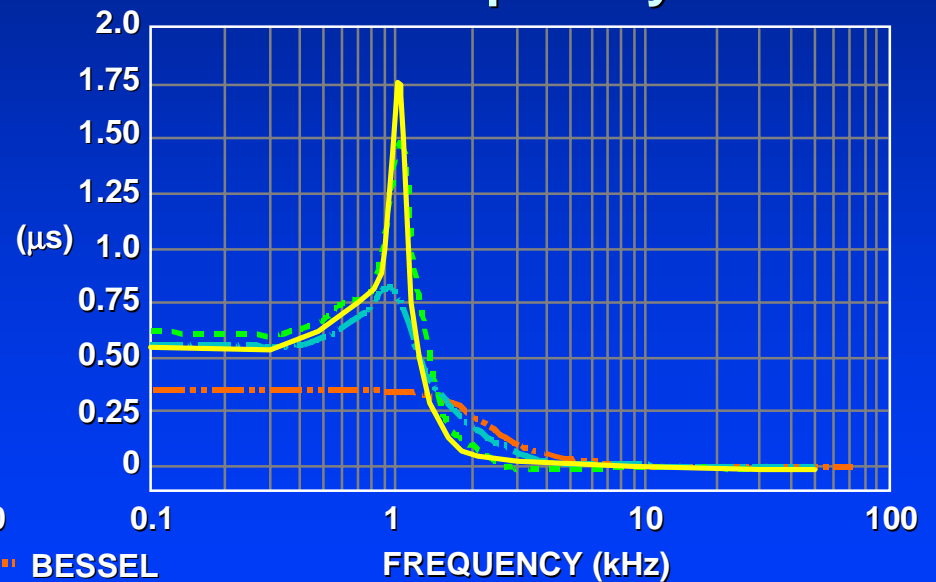
# Filter Response Comparison

- THEORETICAL AMPLITUDE AND GROUP DELAY OF FOUR DIFFERENT FILTERS
- FILTERS ARE 5 POLE
- CUTOFF FREQUENCY NORMALIZED TO 1kHz

### Amplitude



### Group Delay

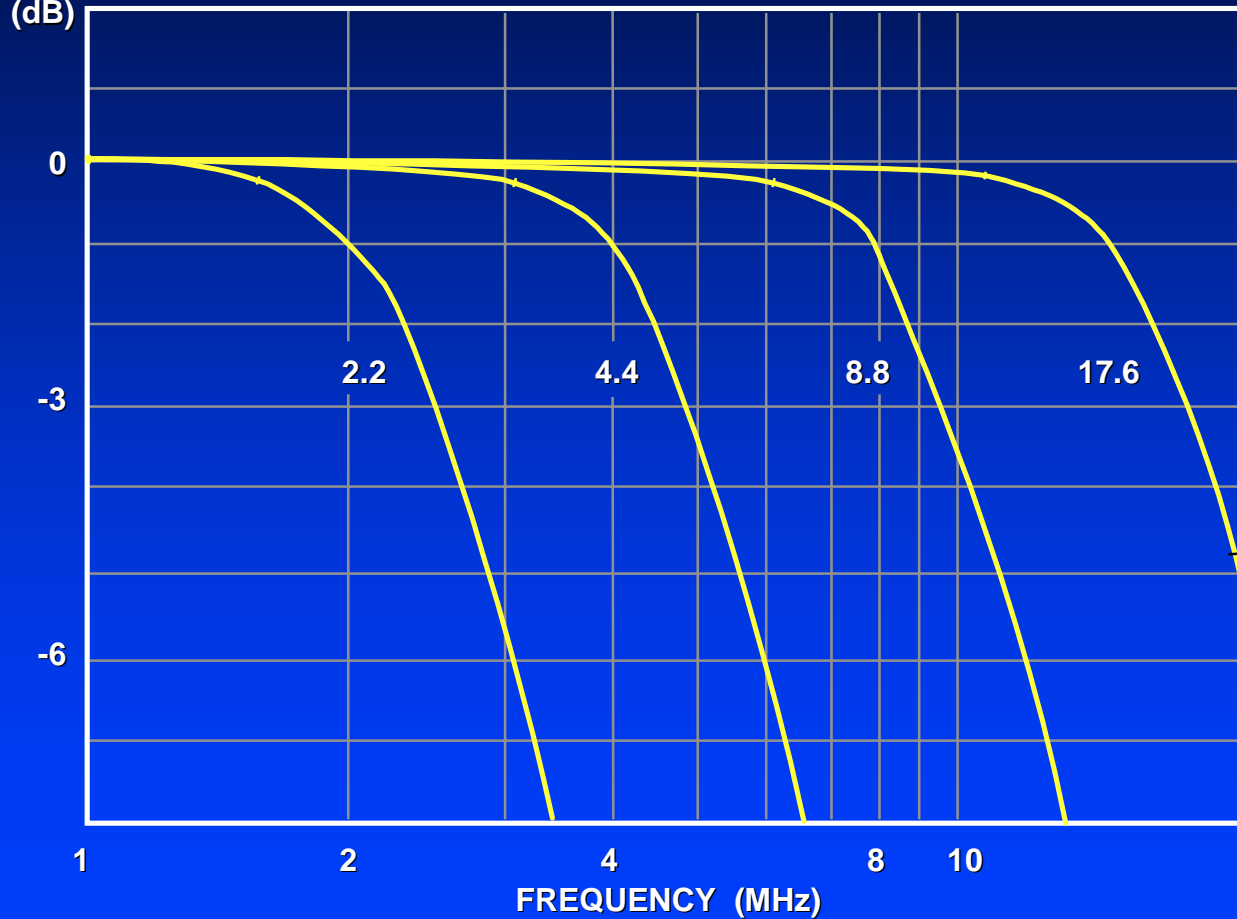


- · · · BESSEL
- · - · BUTTERWORTH
- · - · CHEBYSHEV
- ELLIPTIC



# Filter Cutoff Frequency Selection

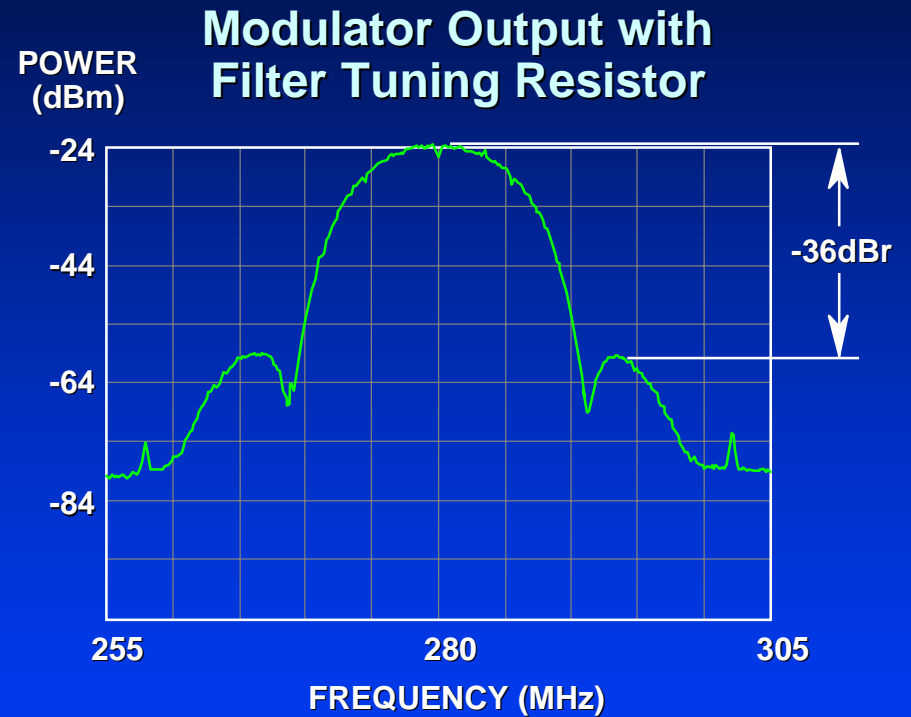
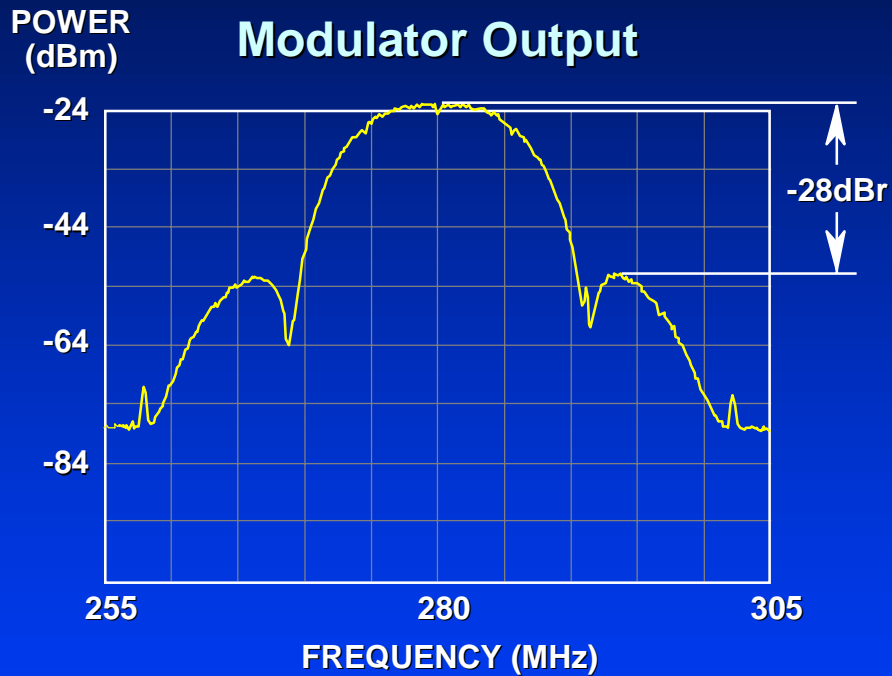
ATTENUATION  
(dB)



SEL 1	SEL 0	$f_{3dB}$ (MHz)
0	0	2.2
0	1	4.4
1	0	8.8
1	1	17.6



# HFA3724 Performance Data



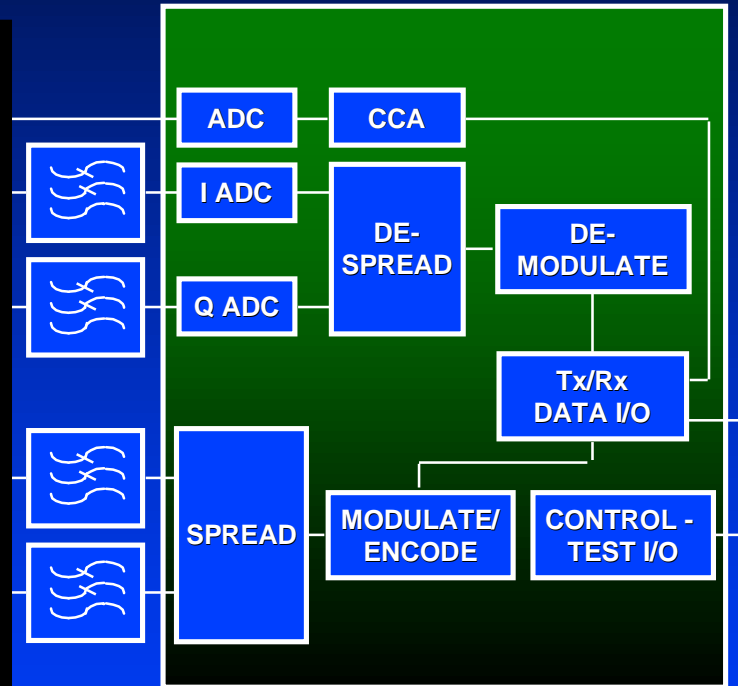


# Baseband Section Issues



- ANALOG TO DIGITAL CONVERSION
- SPREAD / DESPREAD
- MODULATE / DEMODULATE
- CLEAR CHANNEL ASSESSMENT
- DATA FORMATTING AND HEADER CREATION

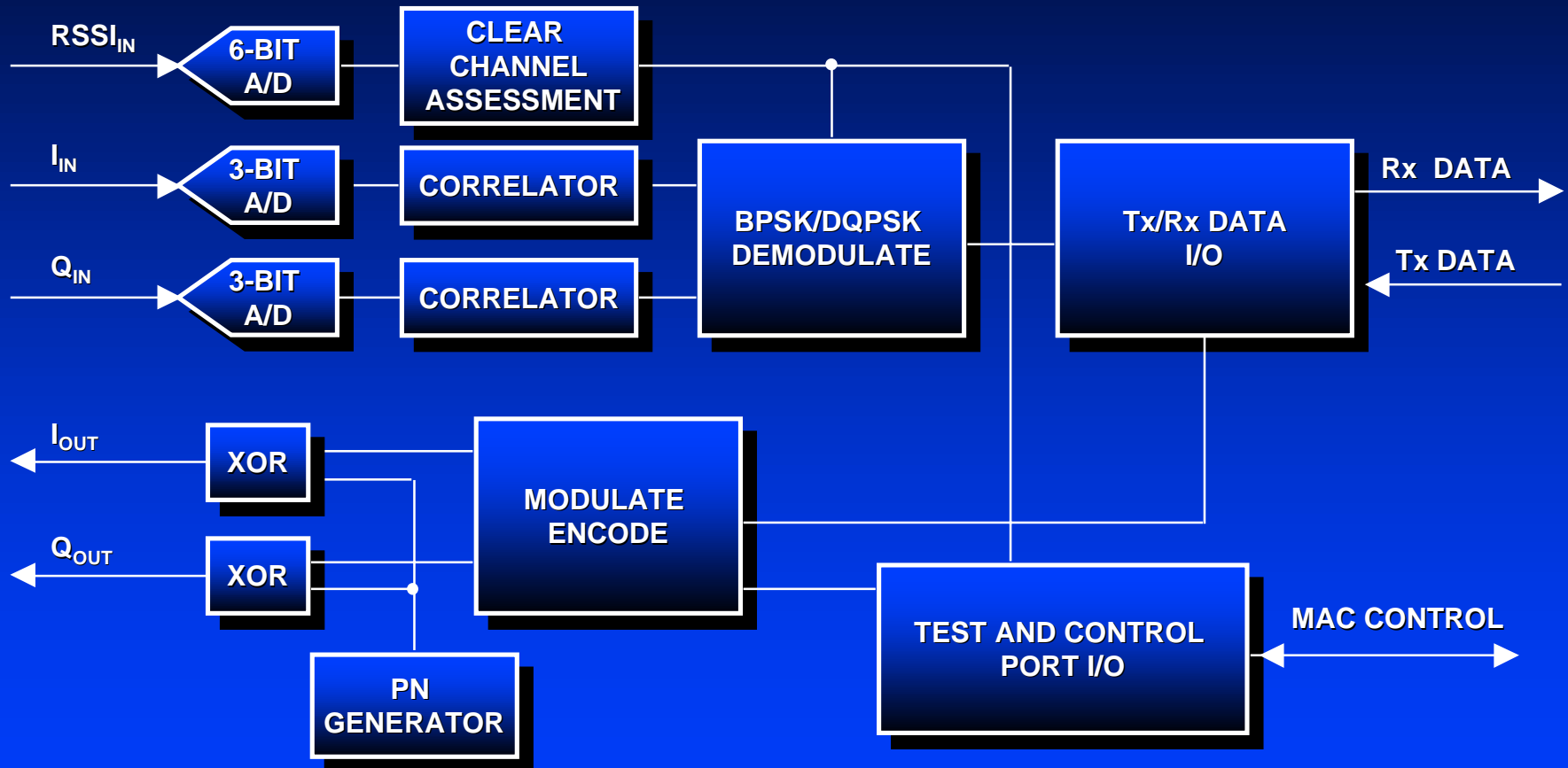
DUAL SYNTHESIZER



HSP3824  
BASEBAND PROCESSOR



# HSP3824 Baseband Processor

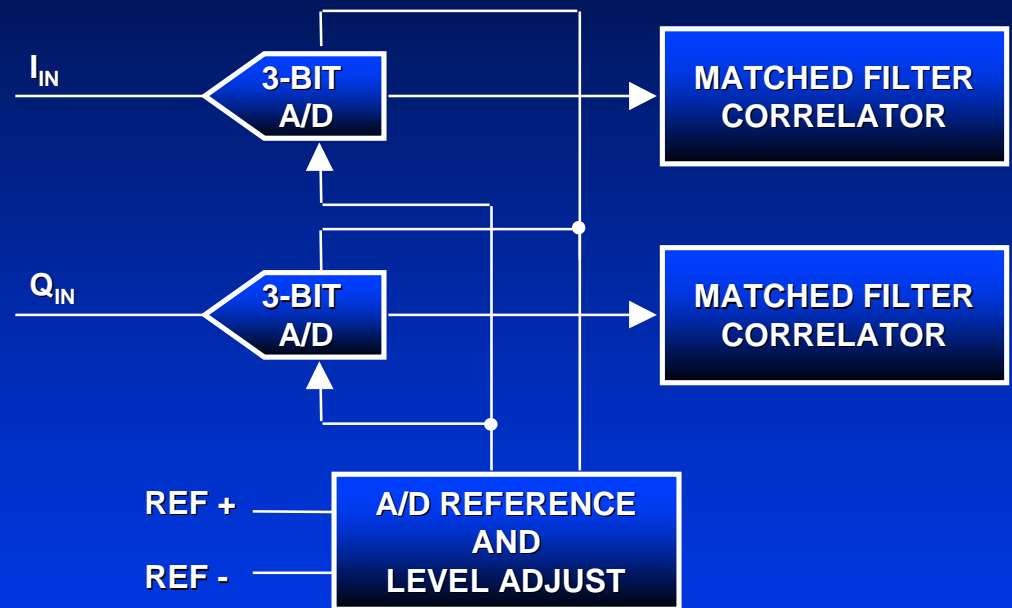






# HSP3824 Analog to Digital Converters

- 44Msps, 3-BIT FLASH A/Ds SAMPLE I/Q INPUTS
- PATENTED CIRCUITRY ACTS AS AGC TO KEEP A/Ds AT FULL SCALE



- 2Msps, 6-BIT A/D SAMPLES RSSI

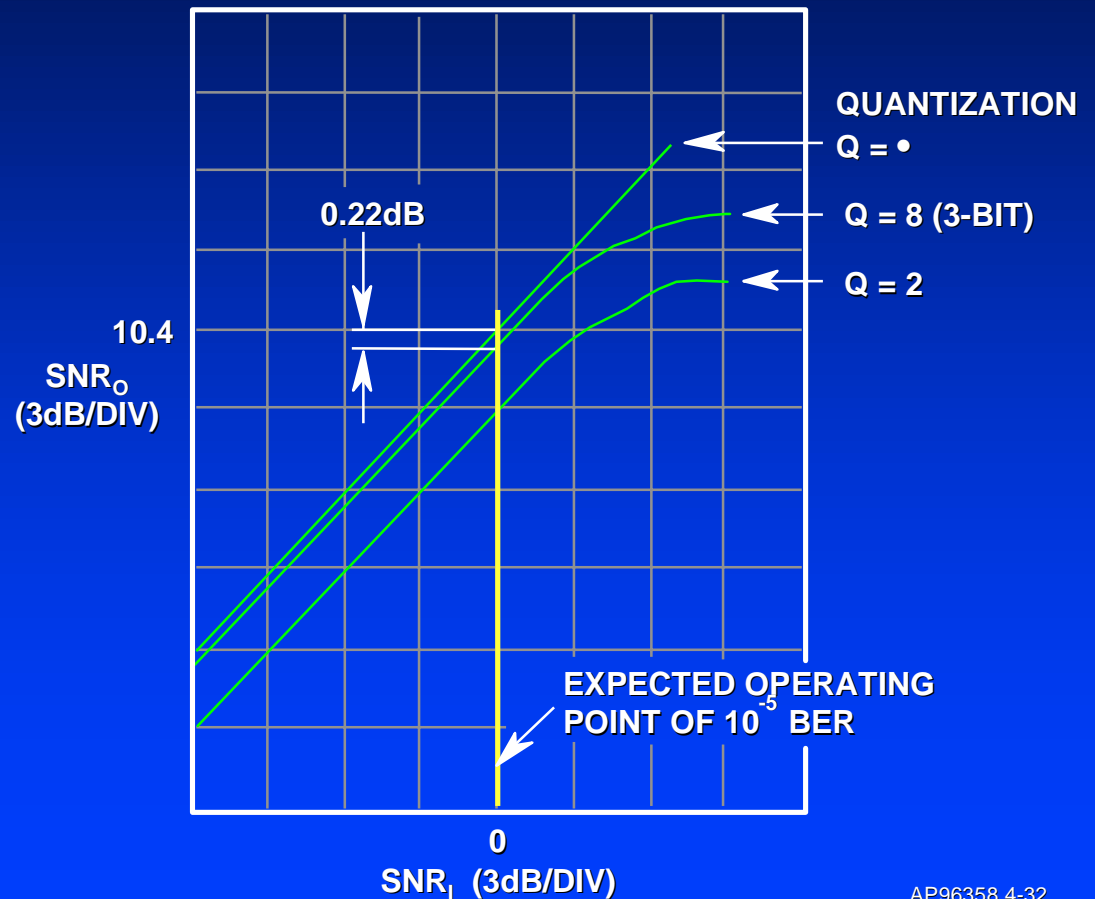




# Why 3-Bits?

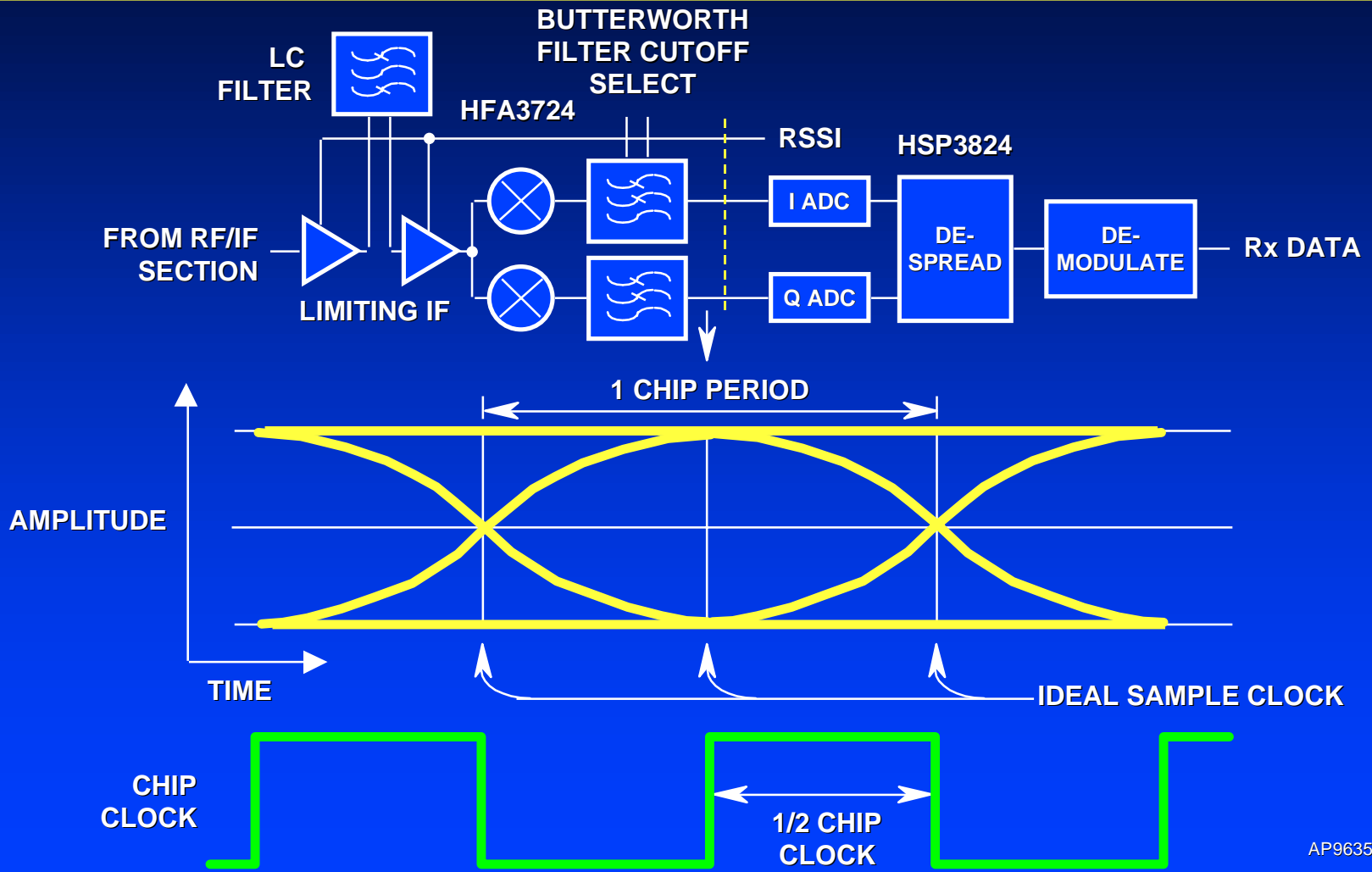
- 3-BIT QUANTIZATION GIVES SIGNIFICANT ADVANTAGE OVER 1-BIT
- QUANTIZATION TO GREATER THAN 3-BITS YIELDS NEGLIGIBLE IMPROVEMENT

Correlator  $SNR_o$  vs  $SNR_i$



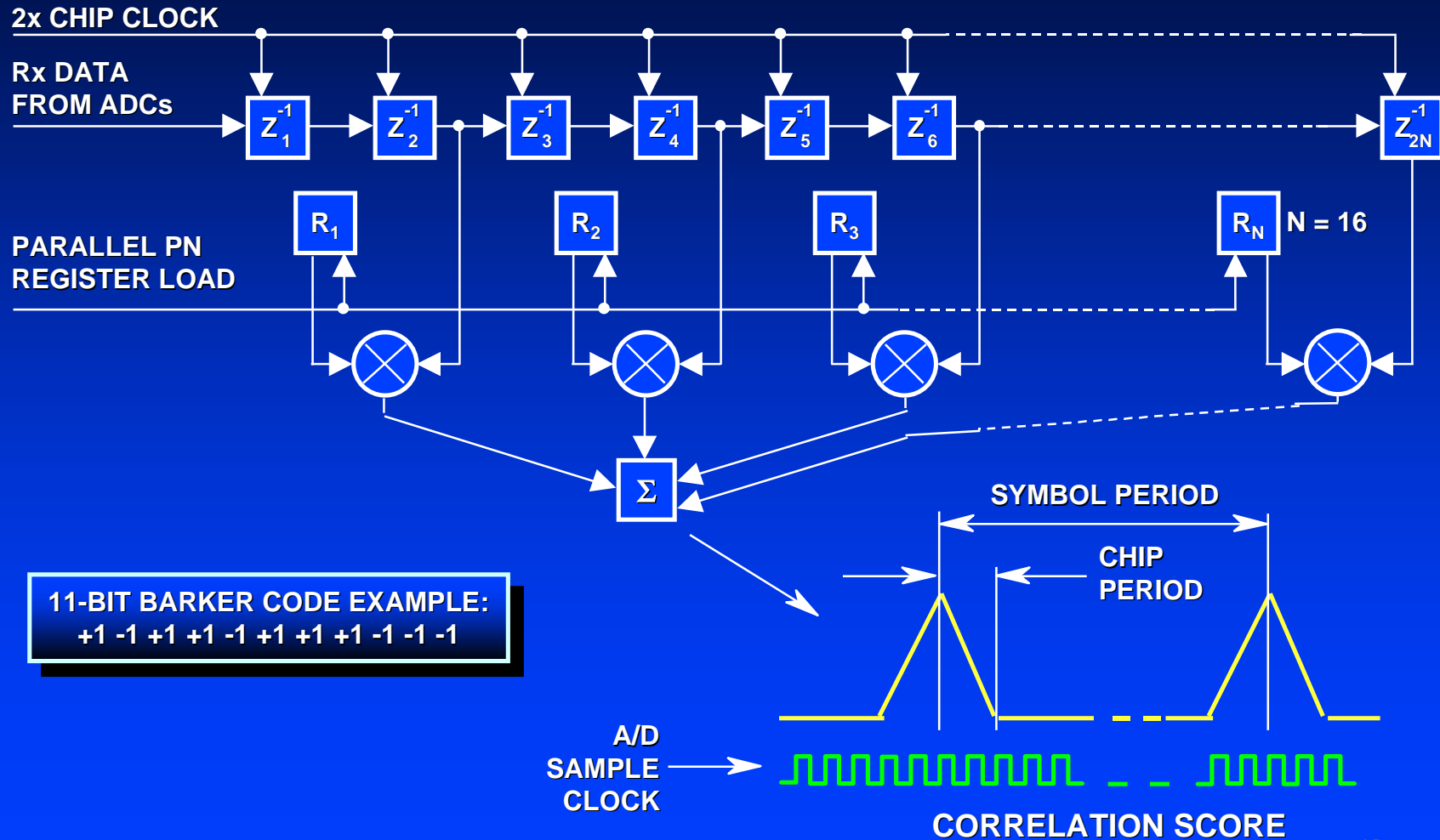


# Sample Clock Timing





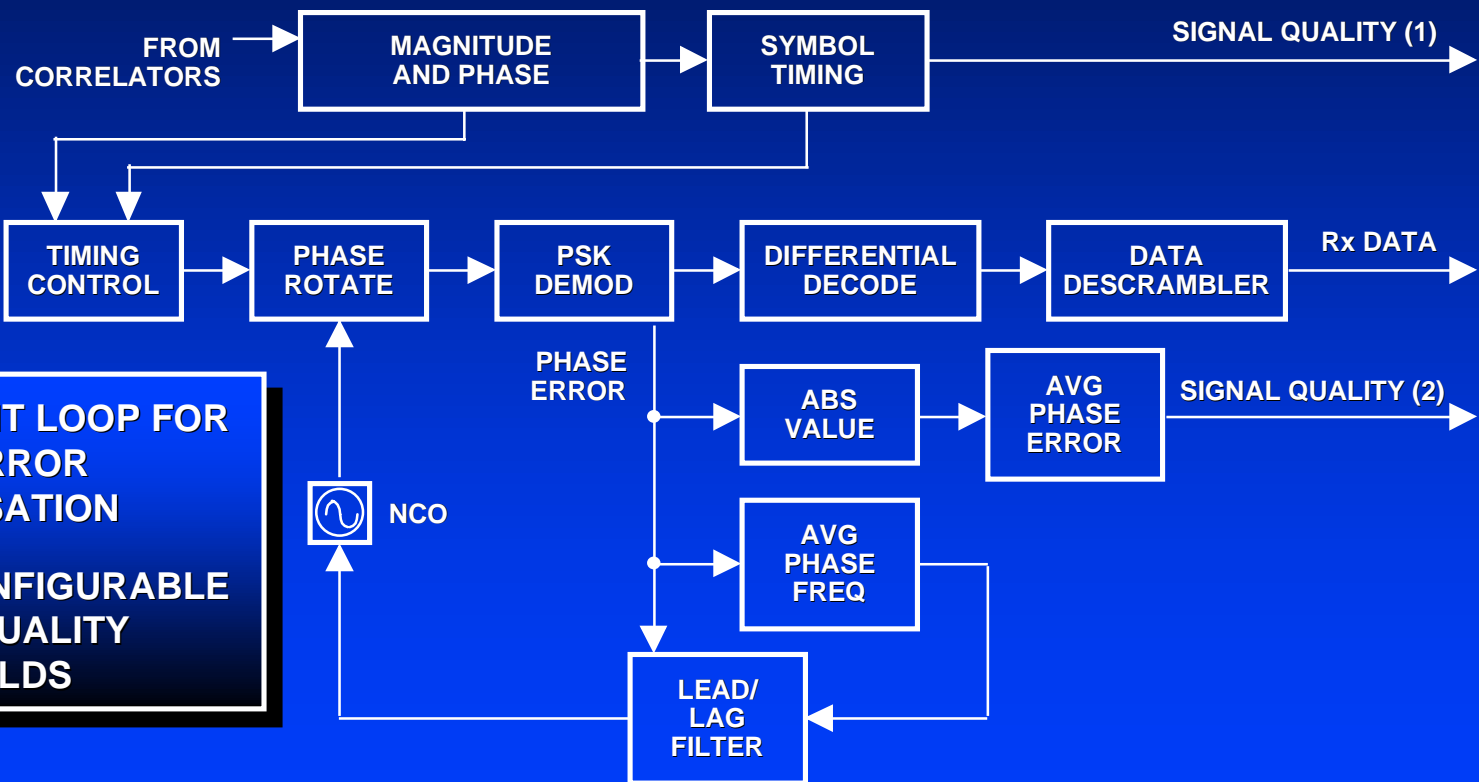
# HSP3824 Matched Filter Correlator





# HSP3824 Receive Section

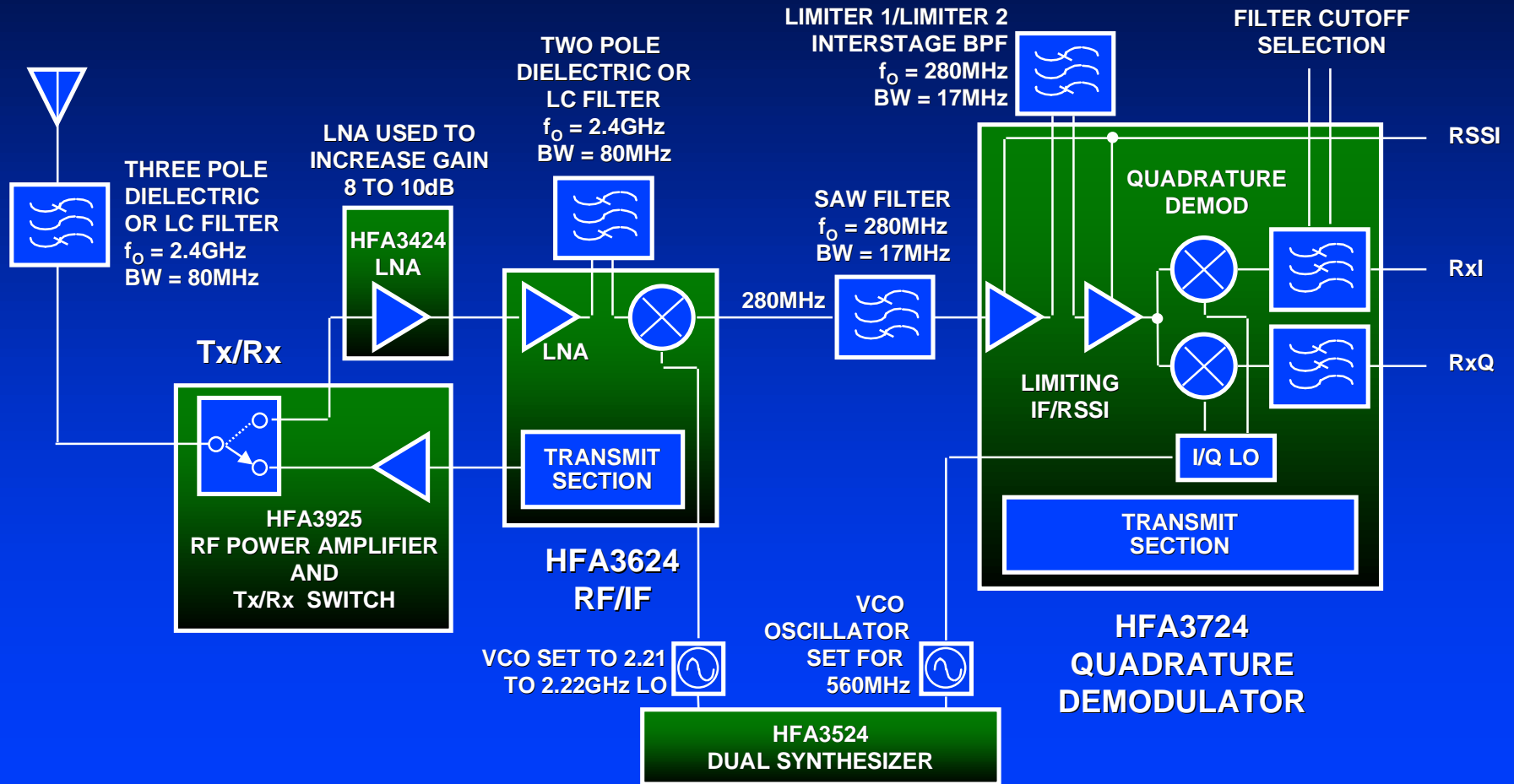
## DPSK Demodulator



- COHERENT LOOP FOR PHASE ERROR COMPENSATION
- USER CONFIGURABLE SIGNAL QUALITY THRESHOLDS

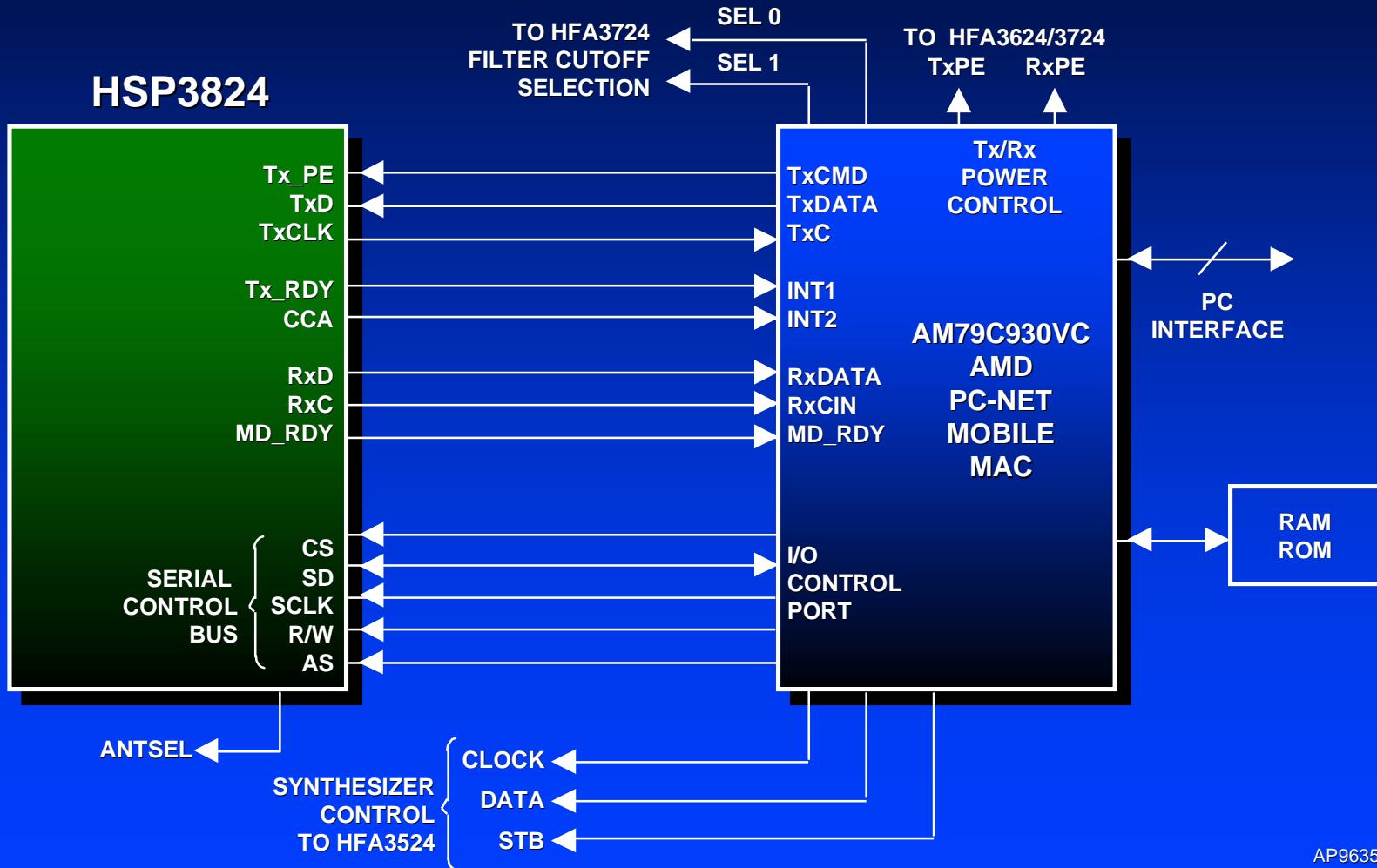


# Complete RF to Baseband Circuit



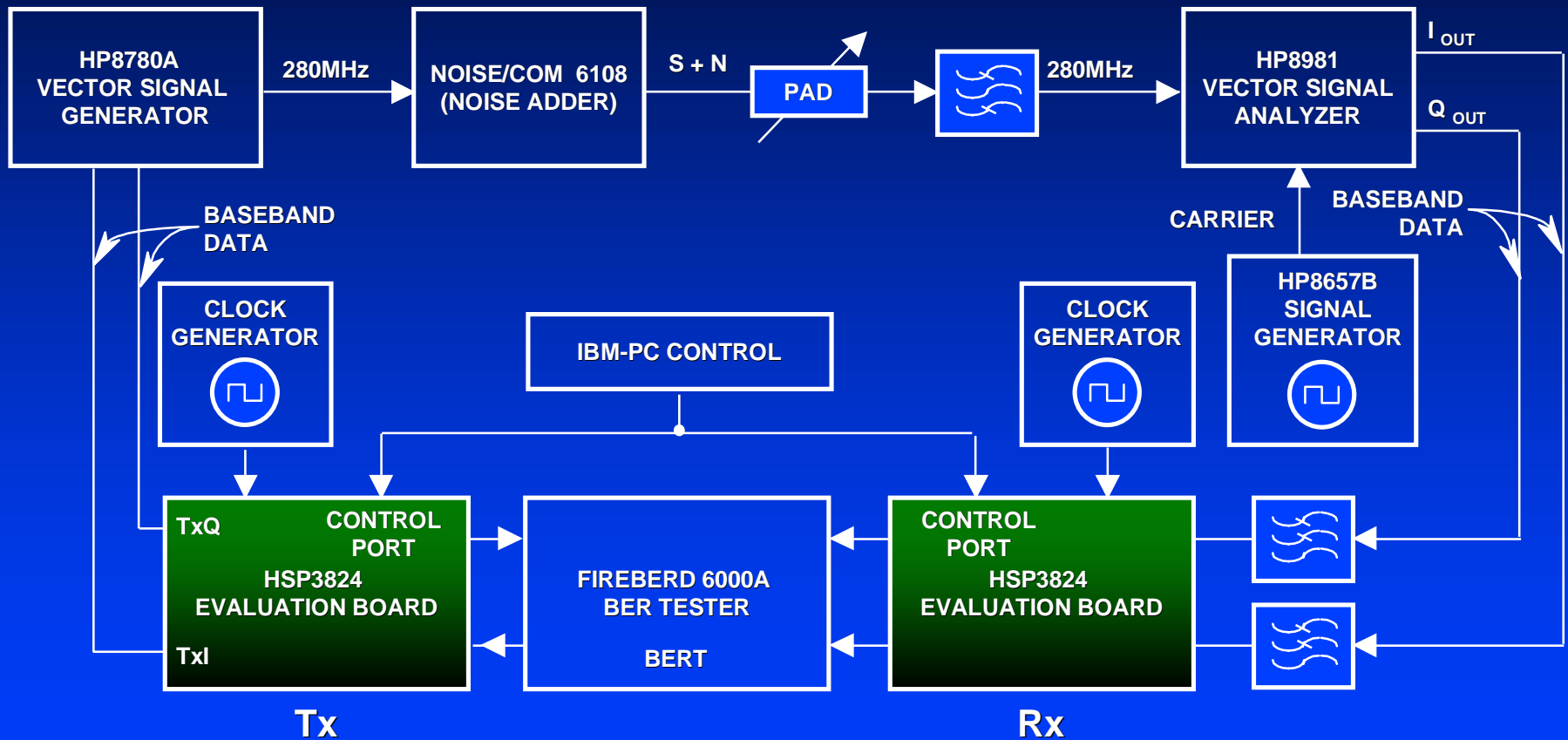


# MAC-PHY Glueless Interface





# HSP3824 Evaluation Setup





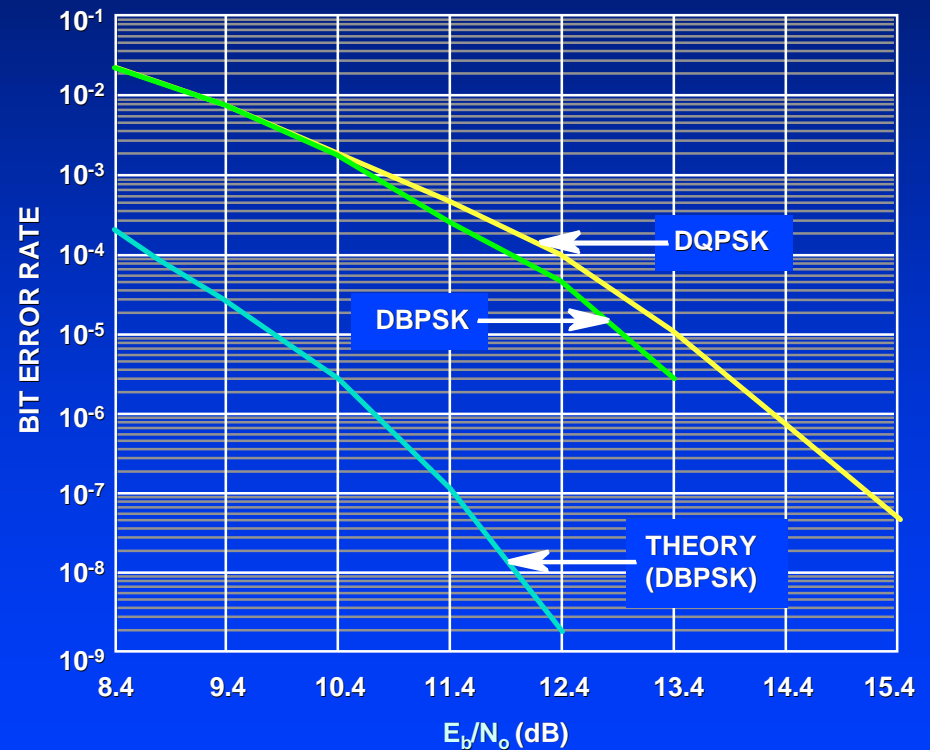


# BER Performance

PARAMETER	LOSS IN dB @ $10^{-5}$ BER
PHASE NOISE	*
LIMITING AMPLIFIERS IN IF	0.5
I/Q PHASE AND AMPLITUDE IMBALANCE	0.75
A/D LINEARITY	NEGLIGIBLE
A/D HEADROOM vs BIAS ADJUSTMENT	NEGLIGIBLE
OSCILLATOR OFFSET IN CORRELATOR AT 50ppm	0.22
SAMPLING STRADDLING OFFSET AND IF + LPF FILTER LOSSES	2.0
QUANTIZATION AND DQPSK DEMOD LOSS	0.5
PHASE LOCK LOOP NOISE	0.1
DESCRAMBLER ERROR EXTENSION	0.5

\* Depends on synthesiser

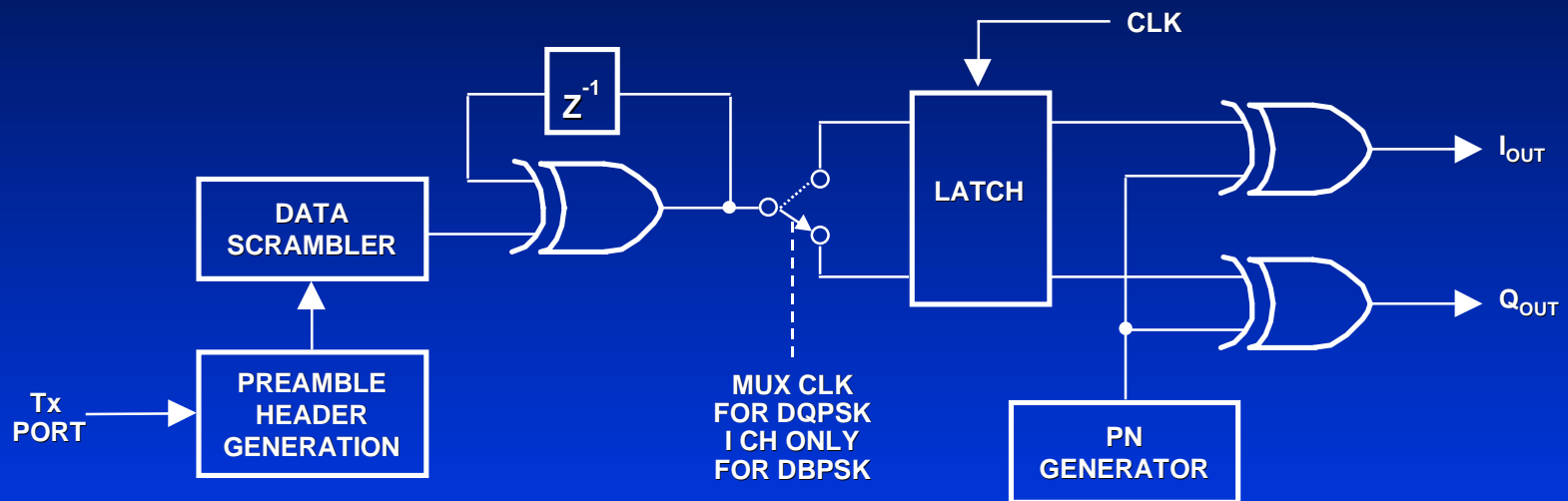
## BER vs $E_b/N_o$ Performance





# HSP3824 Transmit Section

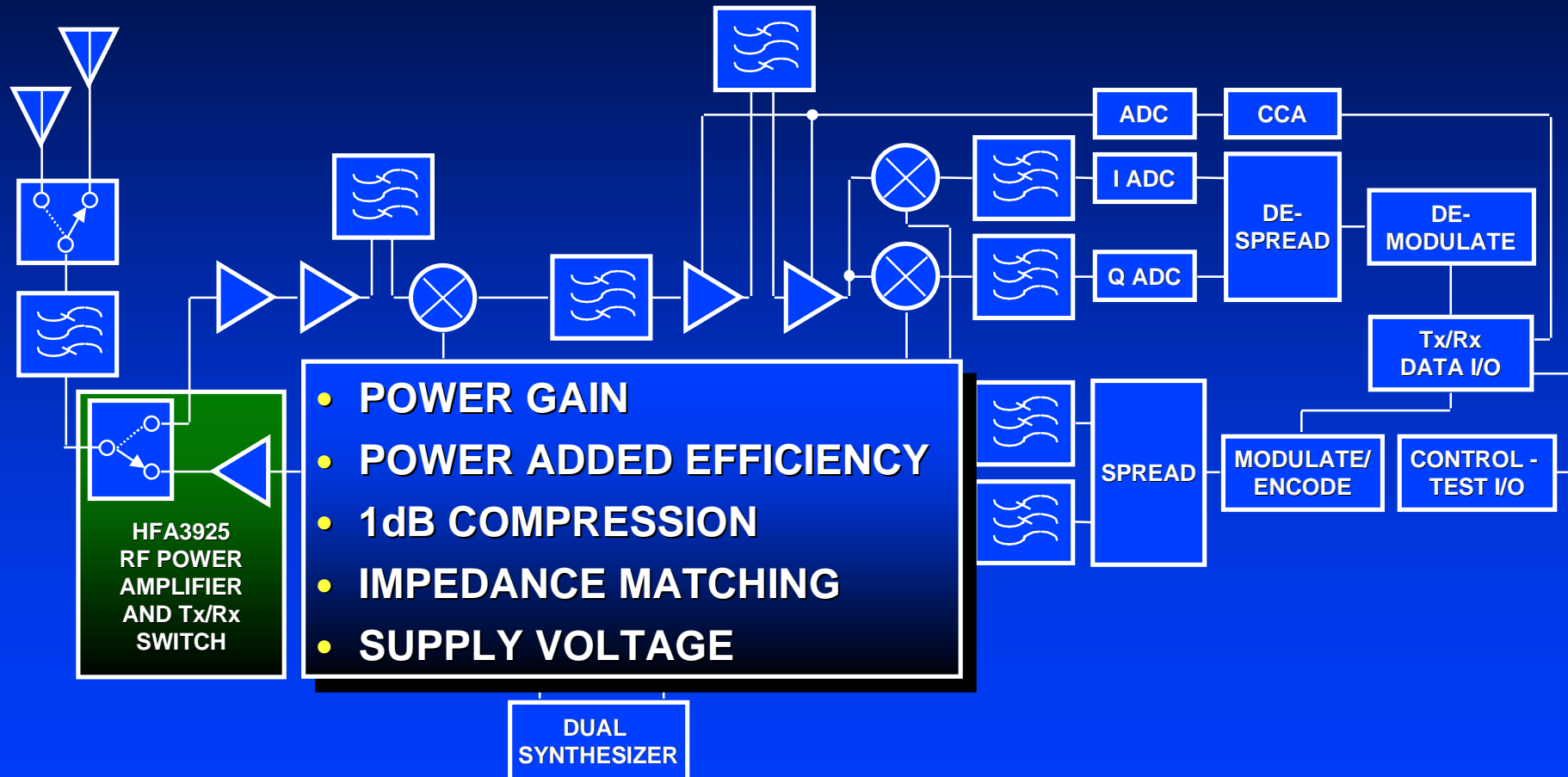
## DPSK Modulator



- FOUR SELECTABLE, INTERNALLY GENERATED PREAMBLE/HEADER FORMATS
- SELECTABLE CHIP SEQUENCES OF 11, 13, 15 OR 16 CHIPS PER SYMBOL
- SYMBOL RATE =  $MCLK / (N * CHIPS \text{ PER SYMBOL})$   
WHERE  $MCLK$  = SAMPLE RATE CLOCK AND  $N$  IS A PROGRAMMABLE DIVIDER OF 2, 4, 8 OR 16

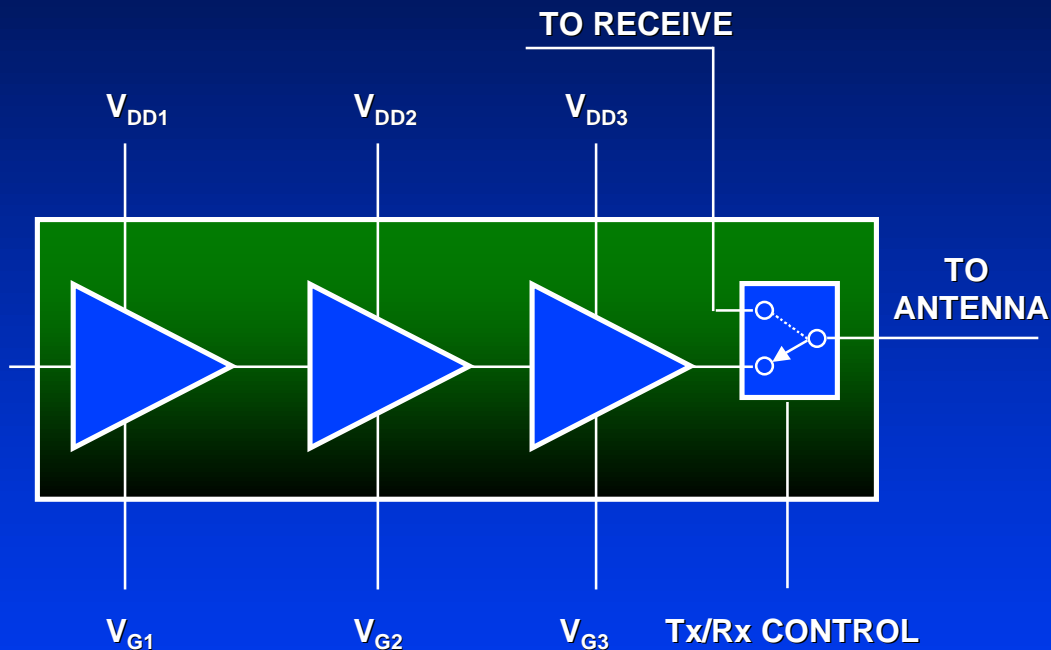


# RF Power Amplifier Issues





# HFA3925 RF Power Amplifier and Switch

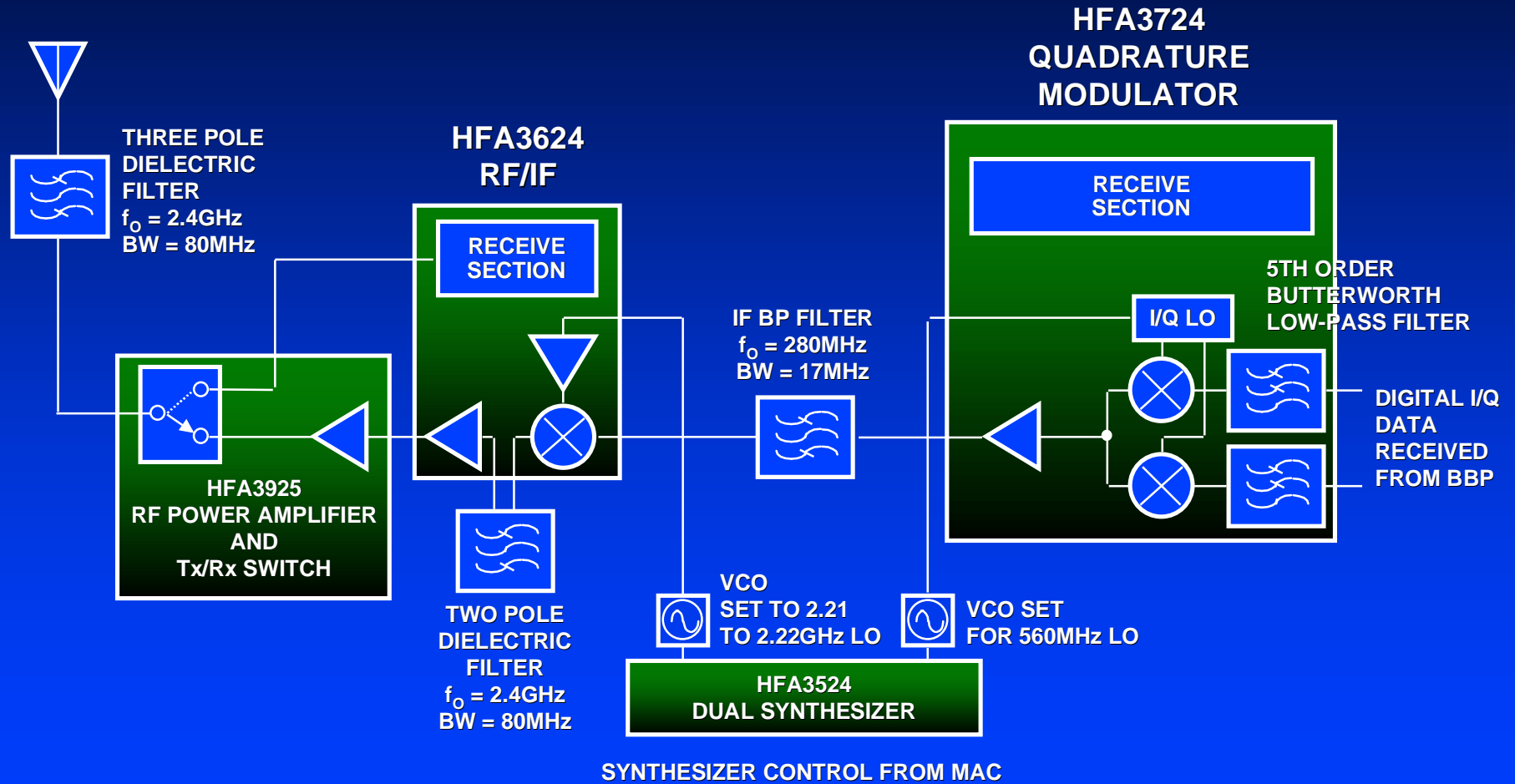


- HIGHLY INTEGRATED GaAs POWER AMPLIFIER WITH T/R SWITCH
- HIGH LINEAR OUTPUT POWER AT OP1dB 24dBm (250mW)
- INDIVIDUAL GATE CONTROL FOR EACH AMPLIFIER STAGE
- VSWR IN/OUT . . . . . 1.75:1
- SUPPLY VOLTAGE . . . 2.7 TO 6V
- T/R SWITCH
  - Insertion Loss . . . . . 1.2dB
  - Isolation . . . . . 12dB

28 Lead SSOP



# Complete Baseband to RF Circuit





# SOFTWARE RADIOS

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- PROGRAMMABLE RADIOS TO ACCOMODATE VARIOUS MODULATION REQUIREMENTS.
- DIGITAL SIGNAL PROCESSING TECHNIQUES.
- DIGITAL FILTERING TECHNIQUES.



# Summary of Cellular and PCS Standards

## IS-95

CDMA/FDM      1.25MHz BW  
20 Chs          BPSK/OQPSK  
798 Users/Ch

## GSM

TDMA/FDM      200kHz BW  
124 Chs          GMSK  
8 Users/Ch

## AMPS

FDMA            30kHz BW  
832 Chs          FM  
1 User/Ch

## IS-54/136

TDMA/FDM      30kHz BW  
832 Chs           $\pi/4$  DQPSK  
3 Users/Ch

What Hardware Works with ALL of These Standards ?

## DECT

TDMA/FDM      1.728MHz BW  
10 Chs          GFSK  
12 Users/Ch

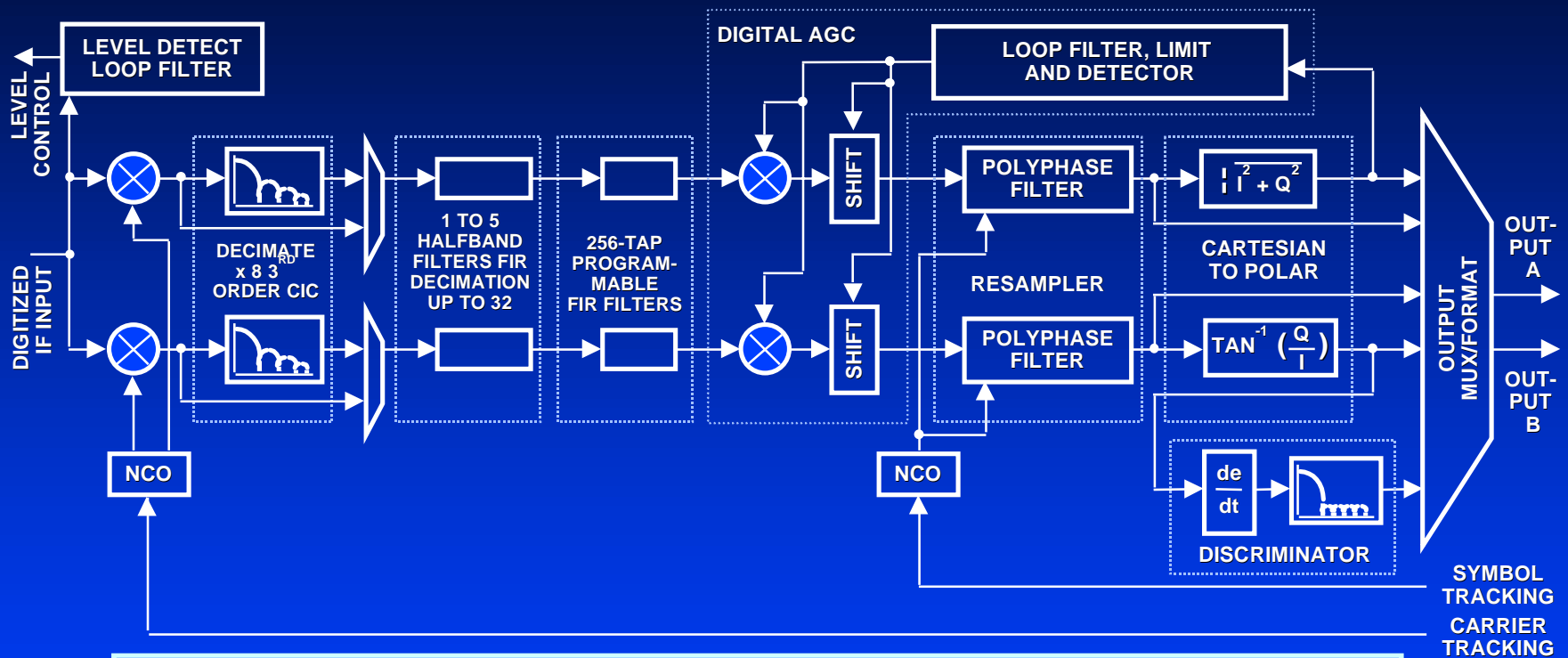
## DCS 1800

TDMA/FDM      200kHz BW  
374 Chs          GMSK  
8 Users/Ch





# HSP50214 - Digital IF Downconverter



## Features

- UP TO 51.2 Msps INPUT TO SUPPORT WIDEBAND CHANNEL SELECTION
- DIGITAL AGC WITH PROGRAMMABLE LIMITS AND SLEW RATE
- OVERALL DECIMATION FACTOR RANGING FROM 1 TO 2048
- SUPPORTS DEMODULATION OF AM, FM, FSK, DPSK